

A PORTABLE SYSTEM FOR
DETECTING INFRASOUND

By

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Abstract: The purpose of this project is to create a device to detect infrasound communication from elephants in order to inform handlers of possible impending aggressive behavior. Elephants often communicate using infrasound which is low-frequency sound below the threshold of human hearing. Elephants may be trying to communicate with zookeepers but the handlers are unable to hear their call. Knowing that an elephant is communicating may give handlers time to move to safety. A device is designed and prototyped that is capable of monitoring an input signal for infrasound and produces a warning alarm for handlers. This device can also record audio for long periods of time to a digital storage device. It can be utilized for other areas of study with some modification. The device is low-cost so it would be able to be procured more easily and in higher quantities than more expensive laboratory monitoring equipment. This device could also be used as an educational and research device for students studying animal behavior in the field and laboratory. Infrasound is not limited to only elephants, but hippopotamuses, rhinoceroses and giraffes also communicate with infrasound. Environmental infrasound from sources such as wind turbines, sonic booms, explosions, tornadoes, and earthquakes can also be monitored. Test results showed that the device accurately recorded low-frequency input signals. The device also was able to detect infrasound frequencies and triggered an alarm.

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CHAPTER I

INTRODUCTION

The purpose of this project is to create a device to detect infrasound from elephants in order to inform handlers of possible impending aggressive behavior. Detection of elephant infrasonic communication with real-time feedback will aid zookeepers in remaining safe during elephant-handling activities. A small and inexpensive system could be employed by elephant keepers as a safety device to warn them of aggressive elephant actions. This device could also be used as an educational and research device for students studying animal behavior in the field and laboratory.

Elephants were chosen as the focus animal for this device for several reasons. Safety is the primary concern of this project, and elephant interactions can be dangerous. Elephants can turn aggressive seemingly without warning, so a device that can make interactions safer is a desired product. In 2010, a Toledo, Ohio zookeeper was critically injured when the elephant he had worked with almost daily for seven years suddenly turned on him with seemingly no warning (1). In 2013, a Springfield, Missouri keeper was killed while feeding an elephant he worked with for over 23 years. (2) The latter zookeeper was a friend of Oklahoma State professor Charles Abramson and was one of the inspirations of this project.

Secondly, elephants, as a popular exhibit animal, are readily available at zoos throughout the world for observation, including the Tulsa and Oklahoma City zoos. This availability also makes them easier to study compared to other animals that are rare and secluded. As a popular attraction to children and many other zoo patrons, increasing the safety of their care ensures a more valuable experience for the observers.

Elephants in the wild are a concern for villages in Asia and Africa. A group of elephants will raid a village with little warning. A system that detects infrasound can provide early warning to humans in the area, giving them a chance to move to safety before the elephants attack.

Lastly, in addition to safety, students can learn more about elephant communication by incorporating the device into the elephant enclosure and to other zoological exhibits in which animals emit infrasound.

Elephant Communication

Elephants utilize many senses for communication – hearing, vision, smell, and touch – as do all social mammals. Many of their sounds are audible to human hearing, especially those produced through their trunks. However, elephants also communicate through very-low-frequency sounds, most often referred to as rumbles. These sounds fall into a range known as infrasound, which is a range of sound below the threshold of audible human hearing, generally defined as below 30 Hz. Therefore, any communication an elephant attempts toward humans in this manner would be unheard. The ability of an elephant keeper to be warned when infrasonic communication is taking place could assist the keeper in remaining safe should the elephant emit warning signals.

Elephants are extremely large and powerful animals, and though human-elephant interactions are often safe, accidents do still occur, often with disastrous results for the humans involved. Though declining since World War II, elephants are still used in Asia as beasts of

burden. Their chores include clearing trees and performing other heavy labor tasks. This requires very close contact with their human handlers, putting the handlers in dangerous situations daily (3). In zoos throughout the world, there are multiple elephant and human accidents yearly. Many of these accidents occur between elephants and keepers that have interacted together without incident for many years, only to have one poor interaction that results in the severe injury or death of a keeper. The deaths mentioned previously are just two of the approximately 500 that occur each year when humans interact with elephants (4). Any system that may warn keepers about impending aggressive behavior would be an improvement and possibly save lives.

Thesis Objective

The objective of this thesis is three-fold. First: to design and construct a low-cost device that is able to monitor infrasound frequencies and warn zookeepers of possibly aggressive behavior by the elephants in their care. Second: to have the ability to record for extended periods infrasound data that can be analyzed in the future. Lastly: to have a device that can serve as an educational tool and a basis for further studies as research equipment is expensive.

The primary function of the device would be to constantly monitor frequencies in the infrasound range and produce an audible alarm to signal zookeepers that the elephant is communicating in infrasound and that they should move to a safer place. This warning system could potentially save lives; therefore, it is the most important objective. However, the information that could be gathered through this device could assist in future elephant communication research, so other functions will be included to assist in that effort. The device should also be able to record the sounds to a storage device for later analysis.

Design Considerations

As a safety device, two requirements need to be established in order for it to be effective. First, the device needs to be small, light-weight, and battery-operated; the availability of electrical outlets and the needed placement of the device would vary between different elephant enclosures, zoos, and facilities, as would the voltages provided in different countries. Though data acquisition systems are readily available for laptop computers, their size and power consumption does not lead them to be used because they do not meet the portability and low-cost objectives. Secondly, it needs to be easy to use with little to no setup required to ensure reliable function during each and every use. A complicated system that requires adjustment or multiple button presses each time the system is powered on could accidentally be configured incorrectly which could result in the alarm function disabled and unable to properly monitor and warn zookeepers.

In addition to safety, another consideration is cost. First, the proposed device will often be used in outdoor animal enclosures. The device could easily be damaged by rain and other environmental causes. Additionally, the elephants themselves could damage the device, no matter how hardened a casing it could possibly have. By using a low-cost device, the device could be replaced easily without much expense, especially compared to a laptop-based system. Secondly, due to the size of enclosures, having multiple warning devices in several locations would increase the ability to properly detect the infrasound signals and ensure that a device was near enough to a keeper to hear the alarm. Lastly, producing the device at low cost would allow researchers to utilize these in poorer areas of the world that cannot afford expensive equipment. As infrasound travels across long distances, villagers could be warned well in advance of an elephant raid, giving them time to seek shelter. Given the right sensor, there is no need for the elephant to be extremely close to trigger the device.

Though designed as a safety device, the system could be easily modified to be used for a variety of applications beyond that of elephant-human interaction safety. Animal behavioral scientists could utilize the devices in many scenarios to record animal sound in nature as well as in laboratory experiments. Commercial recording systems are often expensive. These are out of the budget of public schools, and university systems may have some recording devices, but this device could potentially allow many more students to have access to recording devices that could be used to study animal sounds and behaviors.

Chapter 2 explores the literature involving infrasound measurement and recording, tools for controlling animal behavior experimentation, and elephant calls. This review will explain concepts used in the design of the device. Chapter 3 covers the methodology and design of the device, including the sensor, controller, interface, and software. Chapter 4 reports the test results of the device. Chapter 5 concludes with future work and offers suggestions on how the capabilities of the device can be extended.

CHAPTER II

REVIEW OF LITERATURE

Elephant Listening Project

The study of elephants and their behaviors are a common occurrence in their native lands of Africa and Asia. The Elephant Listening Project (ELP), part of The Cornell Laboratory of Ornithology at Cornell University, has been studying the sounds of elephant calls for over 30 years. Katharine Payne and her team worked mostly with African elephants in western Africa. The fundamental frequencies of the calls they recorded were between 15-35 Hz with pressure levels up to 117 dB at a range of 1 meter from the source (4). These low-frequency sounds can travel over long distances because neither the air nor the ground will attenuate the signals significantly. These long-distance communication methods explain elephant behavior which was not previously understood. Elephant groups separated by several kilometers would simultaneously perform similar movements for several hours. They could also, still separated by kilometers, turn and head directly towards each other. Additionally, though males and females spend most of their time apart from each other, they could find each other easily during mating seasons. The researchers tested these theories by playing back the recorded sounds while observing elephant groups. Elephants seen 2 kilometers away from the sound source would freeze and/or turn and walk towards the source and often respond to these prerecorded calls the majority of the time.

The Elephant Listening Project has recorded over 700,000 hours of sounds in the areas of west-central Africa (5). Self-contained recording devices were designed to record for up to 6 months at a time. One particular study analyzed 5 years of data – from 2007 to 2012. The goal of the analysis was to determine if elephant calls could be detected electronically rather than manually. Manual detection consisted of having a person listen to find the elephant calls on recordings. The recordings used for data analysis were at 16-bit resolution at either a 2000 or 4000 Hz sampling rate. The rumbles they found ranged from 8-180 Hz and lasted from 2 to 8 seconds. The fundamental frequency of the rumble was between 8 and 34 Hz with a median of 15 Hz.

Members of the ELP listened to the recordings and tagged true elephant calls in order to score whether the algorithm correctly identified a true elephant call. The algorithm would divide the recordings into 100 millisecond frames to then score with the likelihood of containing an elephant rumble. All audio was converted to a 2 kHz sampling rate and a short-time Fourier Transform was produced over 1024 points. These spectrographs were split into 20 second segments. Though elephant rumbles are highly varied, they typically have constant power at a specific frequency throughout a rumble. Looking at these frequency and power characteristics, 46 elements were identified in a vector for each time frame.

They found that the first harmonic, $2 \cdot F_0$, which ranged from 20-24 Hz, was more prominent than the fundamental frequency. They were unsure as to the reason for this but suspected the microphones' inability to capture the lower frequencies without loss was the cause. These harmonic features were a factor in their scoring. Image filters were applied to the 20 second frames to look at the horizontal characteristics of the signal to determine how well they matched with elephant calls, as well as to reject noise that came from the hard drive used. Utilizing machine learning algorithms, they were able to detect and classify the sounds found on the recordings. Across about 4000 hours of recordings, the algorithm correctly identified an

elephant call between 80-90% of the time and a false positive rate of less than 9%. When the threshold was lowered to reach a 90% detection rate, the false positive rate exceeded 40%. This analysis was performed on an 8 core Dell laptop with 16 GB of RAM, and 24 hours of recordings were analyzed in 9 minutes.

Similar research was performed in India. The researchers performed a comparable spectral analysis and a neural network classification of the signals, having only 21 elements in their vector. Their results were similar to the ELP project, detecting the signal 90% of the time, but having false positives 30% of the time (6).

Low-cost Recording System

The recordings captured by the ELP utilize autonomous recording devices placed in trees in elephant habitats in Africa. These are specialized devices with long recording time of at least 6 months; therefore, the battery and storage capabilities are great. These often come at great cost. Researchers in Sri Lanka developed a way to record elephant infrasound much more economically (7). Their goal was to develop a recording system that was easy to use and could be made at a fairly low-cost. In their country, as in much of southwest Asia, elephants are common and often used as work animals; therefore, elephant-human conflict is also a common occurrence. According to the Sri Lankan Department of Wildlife Conservation, an average of 150 elephants and 65 people die each year due to these confrontations. Some of these deaths occur from elephant raids on villages. Electric fences, built for village protection, are not able to keep the elephants out. A warning system was needed to detect elephants. Keying in on infrasound communications was the goal of their system.

Their system was based around a PC application that would record the elephant sounds. They needed to develop an analog input, an amplifier, an anti-aliasing filter, and an analog-to-digital converter. The output of the analog to digital converter was then sent to the PC that would

record the digital representation of the elephant rumbles. For the analog sensor, they utilized a speaker which was mounted on a stand and directed toward the elephants at the Dehiwala zoo. The speaker was used instead of a microphone because of its greater sensitivity to lower frequencies and because there was no need for additional circuitry in order to power the device, as a condenser microphone would require. Though no specifications on the speaker are given, pictures of the device show it to be approximately 2-3 inches in diameter. The amplifier they selected was a simple op amp inverting amplifier. The resistances chosen gave the amp a fixed gain of 200 V/V. The anti-aliasing filter used was a second-order Butterworth filter with a cutoff of 100 Hz. The Butterworth was selected to keep a constant gain across the passband without ripples. For the analog to digital converter, a microcontroller was selected that had an onboard 10-bit converter. The microcontroller was then connected to the parallel port on the PC, where a software application recorded the data and provided for a simple graphical interface to allow users to see the frequency components of the recorded sounds. The sampling rate utilized was 7500 Hz.

They tested their device at a zoo and captured signals that contained infrasound characteristics of elephant calls mentioned previously. Fourier transforms of these recordings showed strong results in the 7-15 Hz range with duration of about 7 seconds. Their recording system was also tested on a diesel engine, finding the dominant frequency for it to be below 1 Hz. They showed that their system was able to record infrasound reliably, and the cost for the hardware, from speaker input to microcontroller output (excluding the PC and application) was only \$35.

The Propeller Experimental Controller

Microcontrollers are being utilized in other behavioral laboratory experiments. The Laboratory of Comparative Psychology and Behavioral Biology at Oklahoma State University

experimented with the use of microcontrollers, specifically, the Parallax Propeller, rather than the much more expensive laboratory experiment controllers marketed by several instrument companies. In order to engage students in live animal studies, an easier way to get more students involved was needed. Computer simulations were not providing the learning experience that instructors desired, and the expense, size, and setup of the commercial experiment controllers limited the number of people that could be involved in the live animal experiments (8).

The laboratory found the Propeller easy to setup and program, and it could be easily connected to various sensors and outputs, such as switches, thermometers, lights/light sensors, and audio devices. The Propeller is also capable of generating video signals to see real-time data. They found the Propeller Spin programming language to be the easiest to use of any they had experienced. The Propeller is a multi-core microcontroller, with 8 independent cores. The cores, called “cogs”, allowed several unrelated experiments to be run simultaneously without any interference between them. The ability to utilize already-written modules available at the Parallax Object Exchange, as well as the help from the community forums also at the Parallax website, allowed users to write their programs faster and accomplished tasks that may be beyond their programming skill set.

The portability of the device was also noted. Only a few inches in each dimension, and easily powered by USB, the device can be taken and setup nearly anywhere quite simply. This microcontroller has been used for automatic control and measurements that would be laborious if not impossible to accomplish manually in a reliable fashion. The laboratory went so far as to write experiment controller software that they have made available for other behavior laboratories to use.

CHAPTER III

METHODOLOGY

The design of the device began with identifying components capable of meeting the requirements of a portable, infrasound-detecting, and low-cost device. A microcontroller is required to manage the operations of the device. A signal input device, such as a microphone, is required to capture the infrasound frequencies. That input signal would then need to be amplified and filtered. An analog to digital converter (ADC) is needed to convert the analog signal into a digital representation for the microcontroller to work with. An external storage device is needed in order to store the recordings. A digital to analog converter (DAC) is required in order to take the sampled signal and play it back for the user. Finally, a display and interface control buttons and switches are required to operate the device.

Signal Input

Typical microphones used to record voices or musical instruments have poor low frequency response. Common microphones, such as the Shure SM57 and SM58 (Shure Americas, Niles, IL), have a low frequency response that resides at about 0 dB near 100 Hz and continually drops, with frequency response charts showing a -10 dB loss at 50 Hz (9). The charts end there, as these microphones are intended for human hearing ranges, so there is neither a need to show lower frequencies, nor a need to test microphones to respond to such low frequencies.

Frequency response

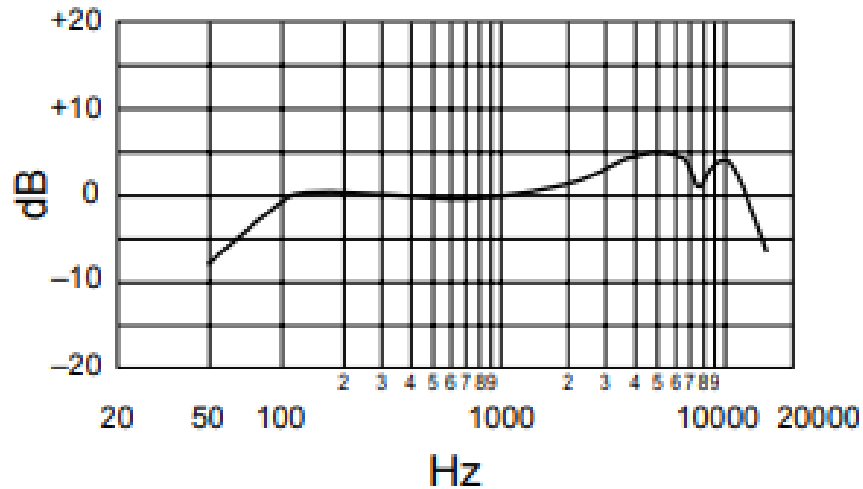


Figure 1- Shure SM58 Frequency Response (9)

Extrapolating from that decline, however, shows that infrasound will not be detected well by these microphones. Clearly a different kind of microphone or sensor will be required in order to detect the infrasound. Additionally, these kinds of vocal microphones cost about \$100, which detracts from the low-cost intent of the device (10).

Specialty microphones designed for infrasound are manufactured. The intended applications for the PCB 378A07 (PCB Piezotronics, Depew, NY) are environmental testing, including wind turbine, sonic boom, explosion, tornado, and earthquake monitoring (11). They are also used to test machinery noise levels, such as industrial equipment and engines, and for the GRAS 41AC-2 (GRAS Sound & Vibration, Holte, Denmark), aircraft and community noise (12). The frequency response for these microphones is very low, some losing only 3 dB at 0.1 Hz. The characteristics of the frequency response are ideal for the purposes of the infrasound recording device.

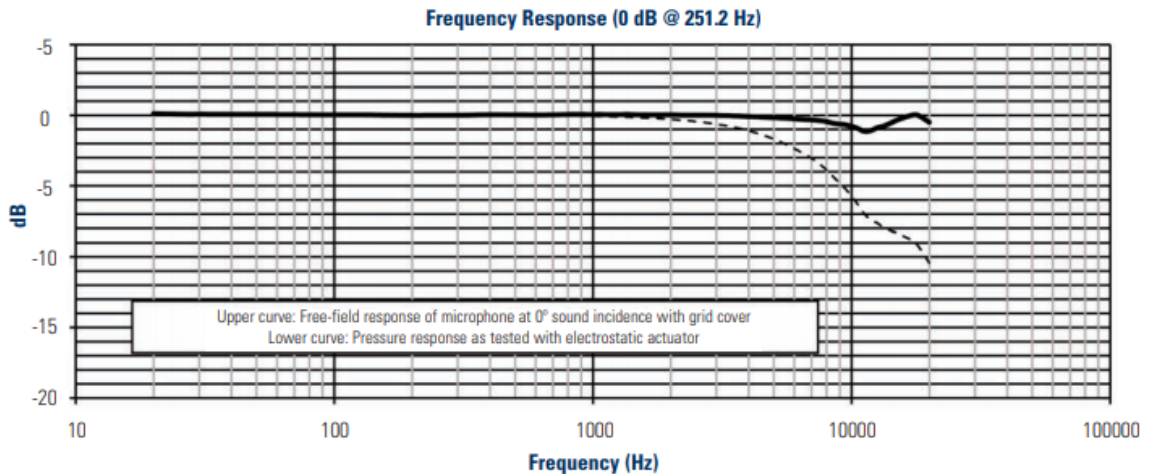


Figure 2 – PCB Piezotronics 378A07 Frequency Response (13)

However, several other characteristics prevent these types of microphones from being used. First, these microphones are often of a free-field design. Free-field microphones work best when pointed directly at the sound source. Sounds coming from other angles are greatly affected. In the operation of this recording device, it is doubtful that a person would be constantly moving the microphone to aim it directly at the elephants, as the most likely need of this device is during movement procedures. Though the effects on this angle of incidence are less for low-frequency signals, this is a factor to be considered. Secondly, these microphones are designed to work with preamplifiers that require at least a 28V power supply (11). That need violates the requirement that this device is small and battery operated. Lastly, these microphones typically cost greater than \$1000 each, and often much more. This also violates the goal of being a low-cost device.

A larger diaphragm in a microphone is better suited to detecting lower frequencies. De Silva and De Zoyza designed their own microphone using a speaker. (7) The underlying concepts between microphones and speakers are the same – sound pressure is converted to electrical signals, and vice-versa. A diaphragm with a coil of wire attached is suspended in a frame. A large magnet is placed behind this coil to induce a magnetic field. When the diaphragm vibrates, the movement of the coils in the magnetic field induces a current in the wire proportional to the vibrations. This concept is feasible and would be sufficient. It does, however, suffer from the

same problem as any microphone in that it needs to be directed toward the sound source in order to be most effective. A second problem that plagues microphones and speakers is wind noise. Wind can cause turbulence at the microphone diaphragm which results in signal generation at very low frequencies (14). These wind-generated signals can be difficult to separate from the actual infrasound intended to be recorded.

Another sensor of a different type is to be considered. Though it is not known with certainty how infrasound signals are detected by elephants, it is suspected that they can detect the signals through their feet (15). This implies that the pressure waves travel through the ground in a seismic fashion. A seismometer is a relatively simple device that can detect vibrations in the earth and convert them to electrical signals. One of the simplest seismographs is very much like the microphone and speaker described previously, but instead of a diaphragm reacting to sound pressure, an inertial mass suspended via a spring reacts to the vibrations. The mass itself is often a magnet, which then induces a current in a coil, effectively turning the vibrations into electrical signals. Unlike a microphone, there is no need to point the seismograph toward the source of the sound, as it is travelling up from the ground into the seismograph, just like it would through elephant's feet.

Seismometers are often used in petroleum exploration. One such item, a Mark Products L-25 (formerly Mark Products, now Sercel, Nantes, France) was available to use for this project. It is a cylindrical device approximately 1.5" in diameter and 2.25" tall, weighing less than a pound.



Figure 3 - Mark Products L-25

No specifications were available for this device, but it is very similar to more recent and available devices. This type of small seismometer is called a geophone. Geophones are used on the ground surface, whereas seismometers are typically much larger and heavier and are buried in order to better sense the very small vibrations of distant earthquakes. Geophones have a spike attached to them which is driven into the ground in order to hold them in place and transfer the ground vibrations to the device.

Due to the lack of specifications, some experimentation was required in order to understand its capabilities. The device was connected to an oscilloscope and then placed on a table. The table was tapped by hand and the measurements recorded. The recorded waveform of an average-strength tap on the table shows that the signal peaks at 50mv and with a frequency of about 29 Hz.

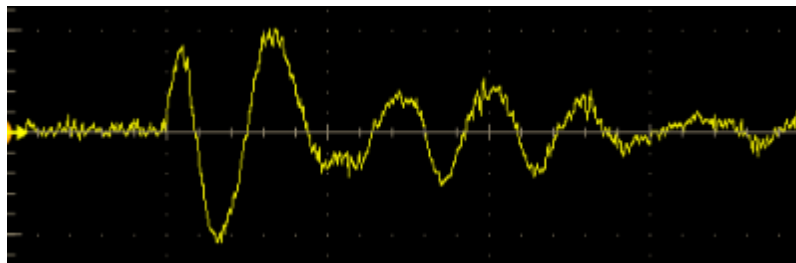


Figure 4 - Signal from L-25 due to tap on table (100 mV/div, 50 ms/div)

The oscillation is due to the nature of the device, with the spring stretching and contracting, dampening to a small amplitude in about 6 periods. Very faint writing on the side of the device was found that looks like it says 28-30 Hz, so this must mean it is the natural frequency of the mass and spring. Specifications for other geophones that are commercially available, such as the Sercel JF-20DX (Sercel, Nantes, France), show that a commonly-used natural frequency for geophones is 28Hz, so it is likely that the L-25 has a similar design (16). Sercel is also formerly known as Mark Products, further increasing that probability. A specification for that device shows that peak output occurs at a frequency of 28 Hz.

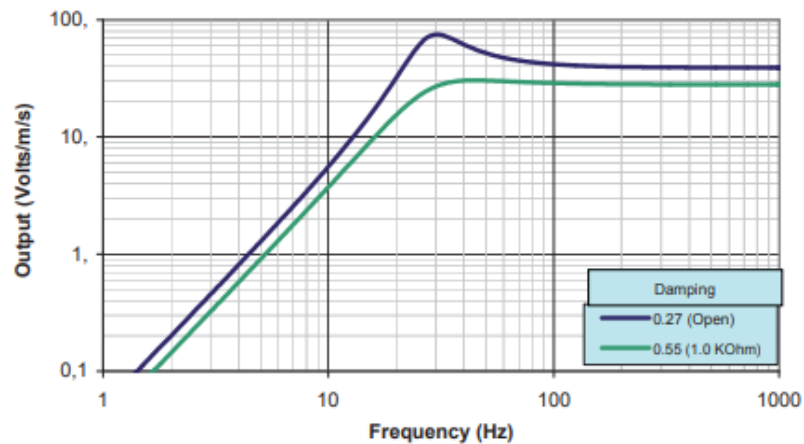


Figure 5 - Sercel JF-20DX 28 Hz Geophone Amplitude Response (16)

Additional testing consisted of patterns of taps on a table to simulate specific frequencies. A 4 Hz table tap (with a stronger tap at each 1-second mark, or every 4 taps, for points of reference) was recorded. The same pattern of dampening oscillations was observed.

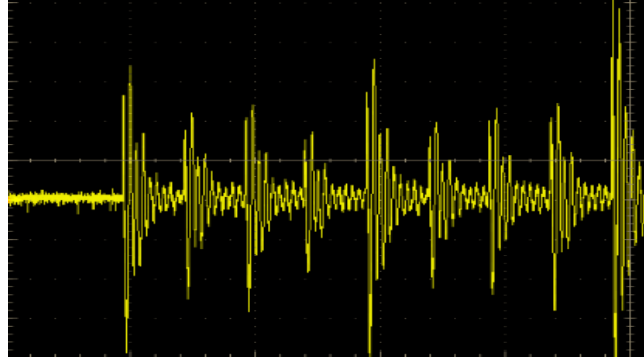


Figure 6 - L-25 response to 4 Hz table taps (10 mV/div, 500 ms/div)

The same 29 Hz oscillations are seen clearly when the 4 Hz signal is zoomed in. It takes until the 5th oscillation for the amplitude to be significantly less than the first oscillation. A method of filtering out the 29 Hz signal could be devised that would leave only the 4 Hz signal.

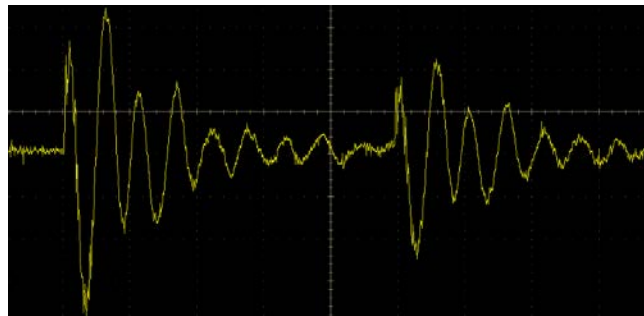


Figure 7 - L-25 response to 4 Hz table taps (10 mV/div, 50 ms/div)

When the tapping rate was increased to about 10 Hz, the signals from the taps were unable to be distinguished from each other. As was seen on the 4 Hz signal, it took until about the 5th oscillation for the amplitude of the signal to be significantly lower than the first oscillation. At 29 Hz, that comes out to be about 116 ms. Therefore, any signal with a period smaller than about 116 ms will be hard to distinguish from the natural oscillations of the spring mechanism.

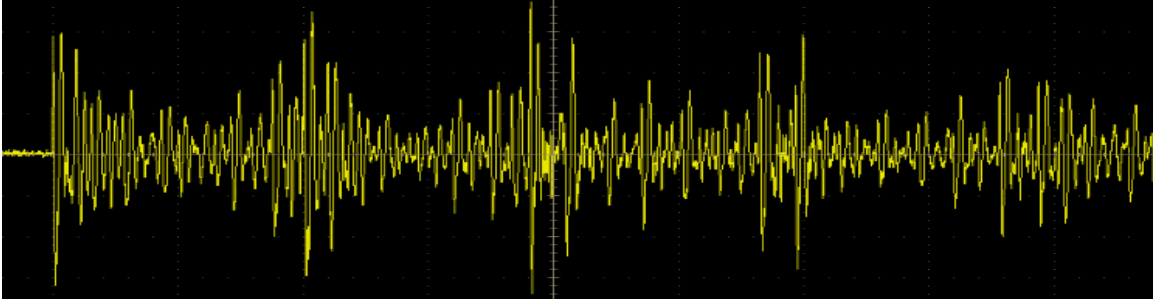


Figure 8 - L-25 response to 10 Hz table taps (30 mV/div, 500 ms/div)

At 10 Hz, the period is 100 ms, and the signals all seemed to mesh together with no way to differentiate between true signal and oscillation. In figure 8, stronger taps were generated at 1 Hz for points of reference. Though those 1 Hz signals are able to be identified, there is no clear 10 Hz signal to be seen under it.

A zoomed-in figure of the same oscilloscope capture shows the difficulty in separating the signals. There is a 1 second difference between the two peaks with the highest amplitude in the figure. Though there are some stronger peaks in between that may be part of the 10 Hz signal, it is just too intertwined to differentiate. The oscillations of the mass have not diminished enough by the time another tap was registered for the new signal to stand out from the oscillations.

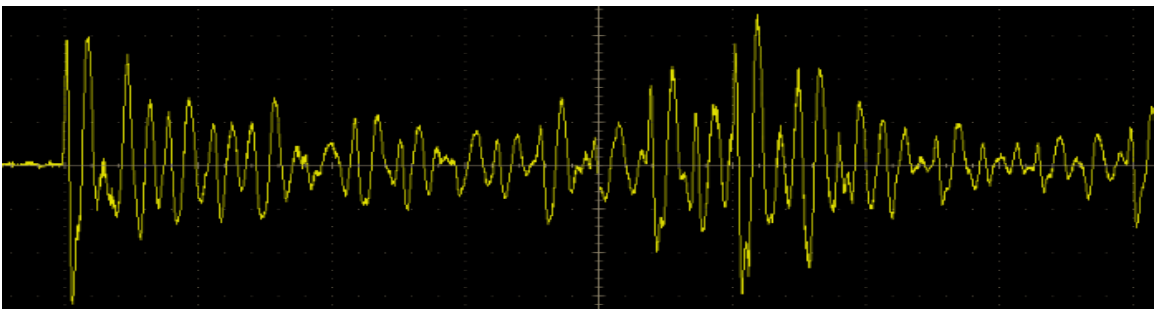


Figure 9 - L-25 response to 10 Hz table taps (30 mV/div, 300 ms/div)

Any infrasound that was picked up by this geophone would be hard to identify specifically by frequency and amplitude for signals above 10 Hz. Therefore, this sensor would not be a good choice for anyone wishing to capture the true audio signal of an elephant call.

However, the sensor may still be effective for simply capturing whether there is any infrasound at all in a location. For the purposes of sensing infrasound as a safety device, this geophone type of sensor may be sufficient for triggering an alarm at the presence of an elephant call.

Microcontroller

The primary component of this device is the microcontroller. Selecting the microcontroller required estimating the microcontroller speed needed, the core bit size, and the amount of IO pins required. As this device is intended to be low cost and small, a microcontroller would be the most likely solution to meet those requirements. The clock frequency required is tied to the highest sampling speed desired. The goal of this device is to sample infrasound, which is less than 30 Hz, so a slow microcontroller would be sufficient. However, the additional objective of making this device usable in many applications requires a much higher sampling rate, such as the CD-quality sampling rate of 44,100 Hz. Each sample will take dozens of clocks for the operations, not to mention the additional processes of storing the data to an external storage device and any other device functions occurring simultaneously, such as human interface monitoring. A “safe” clock rate would be at least 1000 clocks per sample, therefore a microcontroller clock greater than 44MHz would be the best bet in order to meet the required sampling speed.

The core size of the microcontroller would need to be sufficient to handle the bit depth of the analog to digital converter. An 8-bit core could only handle a -128/127 value from the ADC which would only provide a signal to quantization noise ratio of 48.16 dB. A greater bit depth is desired, so the core of the microcontroller should be more than 8 bits.

The microcontroller also requires enough IO pins to interface with all peripherals. Connections need to be made to at least the ADC, DAC, display, interface buttons, and external storage device. This does not even include any spare pins that should be made available to future

uses or expansion. These requirements allow for no less than 16 IO pins be available on the microcontroller.

There is a wide range of microcontrollers that meets these requirements. Any one of them would be sufficient for the purposes of this device. An additional factor that would help the choice of microcontroller is the programming language. As these devices are intended to be used by individuals who are not necessarily frequent programmers, utilizing a programming language and interface that is easier to understand would make the device more usable by a wider range of individuals. Though some functions may reside in assembly-level instructions, the portions of the code that may need to be changed by future users in order to meet their application should be in a higher-level language that is much easier to access.

To that end, the Parallax Propeller microcontroller meets the desired requirements. In addition to meeting all of the technical requirements stated above, the company has its own programming language called Spin made just for the Propeller. There is also the Propeller Assembly (PASM) language which allows for more detailed control of the microcontroller. This microcontroller has already been used in the Comparative Psychology and Behavioral Biology Laboratory at Oklahoma State University. Dr. Charles Abramson suggested this microcontroller due to its ease of use in his laboratory. Psychology students are not focused on programming yet have successfully used the Propeller to control experiments in their laboratories. Utilizing a Propeller microcontroller for this device increases the likelihood of maintainability and easier modification by individuals who are not strong programmers.

The Propeller is a multiprocessor microcontroller. It contains 8 individual processors called cogs. The starting and stopping of the cogs is controlled by a hub which manages all operations. The cogs process in parallel, but access to main RAM takes place in a round-robin fashion, allotting each cog exclusive access on a regular cycle. (17 p. 21) All cogs have access to

the IO pins and the system clock at the same time. These characteristics allow the Propeller to be run without the need for interrupts to control event-handling. Different cogs can be free to run their dedicated function, such as ADC processing, writing to external storage, display, etc. The processor speed can be multiplied to 80 MHz using a phase-locked loop method.

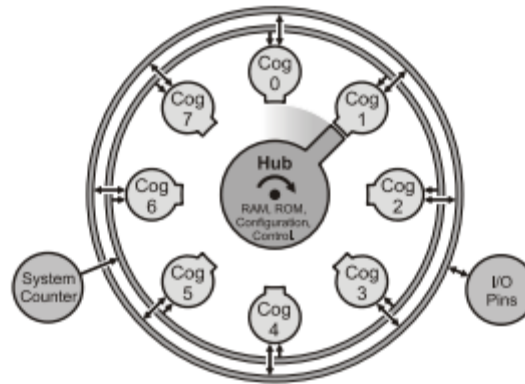


Figure 10 - Propeller Hub and Cog Interaction (17)

The Propeller can be purchased by itself, but also comes in a variety of development boards that contain voltage regulation, USB connections for programming and terminal access, and other peripherals. Determining which development board to use, if any, required selecting what other components were required in order to meet the device requirements.

Amplifier

No matter the sensor type, the input signal would need to be amplified and/or buffered. The signal coming from the input sensor will be AC in nature and in the millivolt range. This signal will need to be adjusted to fit within the range of the analog to digital converter. This range is approximately 0-5 volts. Many amplifying circuits are capable of meeting this requirement. A non-inverting op amp circuit was chosen to accomplish this. The op amp is also powered with 0 and 5 volts. The input signal is sent through a coupling capacitor and biased to the midpoint of the range (2.5VDC.) The circuit also incorporates a potentiometer to allow for a variable gain. This allows the device to accommodate a wide range of input sensors. The design

for this amplifier is modified from the MPLAB Starter Kit from Microchip Technology (18 p. 43). The gain of the amplifier is 2-229 V/V. The end result of this amplifier is a signal with a maximum allowable amplitude of 2.5 VAC centered at 2.5 VDC.

An anti-aliasing filter is required prior to sending the signal to the ADC. The filter is a low-pass filter than has a cutoff less than half of the sampling frequency. Because the signals in question are less than 30 Hz, the Nyquist frequency could be as low as 60 Hz. However, to prevent any loss of signal, the cutoff frequency should be a bit higher than the highest frequency desired to be captured so that is it not attenuated. To minimize ripple in the passband, a Butterworth filter was selected. The frequency roll off of a Butterworth is slow. Therefore, the sampling frequency would need to be higher than double the frequency where magnitude is deemed filtered sufficiently to ensure there is no aliasing.

Analog to Digital Converter

Similar to the point made in the microcontroller section, an 8-bit ADC would not produce a signal with the precision easily capable with common components, nor with the fidelity desired in order to capture a signal faithful to the original. Those 8 bits could only produce 256 points of quantization. Therefore, a higher-order ADC is desired.

ADCs typically come in the range of 8 to 24 bits of resolution. As 8 bits has been deemed too few bits, 16 bits is then considered. Sixteen bits provides for 65,536 points of quantization. If the maximum range of this device of 5 volts, the least significant bit would represent around 0.076 millivolts. This is much smaller than the device would need to differentiate, so that depth of resolution is not needed. Midway between the two, a 12-bit ADC, provides 4096 points of quantization, which results in 1.2 millivolts at the least significant bit. This number is more in line with the capabilities of the rest of the device.

Interface

The device requires an interface in order to control the functions to be executed. A display is required for visual feedback, as well as buttons for control. The number of top-level functions is fairly small, e.g. Monitor, Record, Playback; therefore, a tiered menu-style control setup suffices to completely control the device. Four momentary switches for control are all that is needed (Back, Up, Down, and Select.) This limits some kinds of input, such as character and number input, but keeps the interface simple.

The display chosen is a Parallax 2x16 character LCD (#27922) with backlight and a piezospeaker (19). This display is small enough to meet the small design requirement of the device yet provides enough information to properly use the device. Being a Parallax product, it also interfaces well with the Propeller, so controlling the display is simple. The speaker is loud enough that it can be used to create an alarm to warn zookeepers to infrasound detection.

A 5-LED block is used to show the audio level coming into the device (PN SSA-LXB525-G2YAID, Lumex.) It is similar to many used in audio meter devices. Three LEDs are green, yellow, amber, and red. The input into the ADC is analyzed and the LEDs are lit depending on the maximum value detected over a period of time. This feedback is needed to help the user in setting the gain level on the potentiometer as there is no audio feedback available to inform the user as to the audio level coming into the device. Though the maximum value for a 12-bit DAC is 4095 bits, the input signal needs to be biased at the midpoint to avoid the incoming audio signal from being cutoff with negative-peaked signals. Therefore, when deciding how to light the LEDs, the minimum value is 2047 and the maximum is 4095. The LEDs are programmed to light at 20%, 40%, 60%, 80%, and 90% of maximum value, respectively from green to red.

The device would need to be battery-powered. Optionally, an external DC adapter could also be considered if a power source were available where the device is to be placed.

External Storage

An external storage device must be included in order to capture the recorded audio. This memory should be non-volatile so that the memory is retained when powered down. It also should be low power, and does not need to be excessively large, as these audio files will rarely exceed several megabytes. Raw data audio file types, such as the WAV format, perform no compression on the audio data, and so each sample is stored exactly as it is measured. The WAV format has 40 bytes of header information; the remaining data is all audio samples. Therefore, the file size is: 40 bytes + # of bytes per sample \times # of samples (20). Assuming a 32-bit microcontroller, the largest file that could be handled would be 2^{31} bytes, or 2GB. A 2-byte data word sampled at 10,000 Hz could be as long as 107,374 seconds, which is 29 hours and 49 minutes. Elephant calls last less than 10 seconds, and the sampling rate could be much lower due to the low frequencies being recorded. Theoretically, a single file could record continuously for nearly two weeks at 900 Hz. There is clearly no need for large external storage devices. 2GB of storage would be more than enough.

Digital to Analog Converter

A Digital to Analog Converter (DAC) was included in order to have playback of recordings. The DAC selected is a 12-bit DAC that essentially reverses the process that the ADC and recording procedures executed. The data is read from the SD card WAV files and buffers are filled, with the data being sent to the DAC at the rate specified in the WAV file header. Because the output of the DAC is in the range 0-5 VDC, the output is connected to the headphone jack via a coupling capacitor to block DC current so that only the AC component transfers to the headphones.

Because infrasound is below human hearing thresholds, a simple playback of the WAV file is not helpful if desiring to hear the recorded elephant call. The signal would need to be modified in order for the user to be able to hear any infrasound that had been recorded. A simple software solution to this problem is to provide a method for the user to increase the speed of the playback by lowering the number of clocks between inputs to the DAC. A ten-fold increase in the playback speed would allow an inaudible signal to be heard, e.g., 15 Hz output is 150 Hz.

Software

The Propeller Spin and PASM languages are object-based, meaning that different functions can be separated into modules whose methods can then be called from a higher-level module, the highest level of which is called the top object. Many objects for performing particular functions, such as serial terminal control, SD card file handling, etc., are available on the Parallax website in a section called the Object Exchange (obex.parallax.com.) The exchange is a community-supported, “open source” depository of code written by users and by Parallax employees. It is a relatively small collection of only about 800 objects. These objects range from simple modules as listed above, that are not stand-alone programs themselves, to complete programs that perform various functions, such as an IR Remote Decoder, a function generator, and a pulse-width modulated motor driver. These objects are programmed in Spin, PASM, and some in C. Perhaps even more helpful are the forums where comments and questions can be posted and other users and experts can answer questions and help with debugging code. During the course of this project, the members of the forum were extremely helpful in my attempts to understand the Propeller and how to use it to accomplish my goals with it.

Many of the objects available were written to function with specific development boards from Parallax. Because of that, most modules require modification in order to utilize the code for development boards of a different configuration. Sometimes the changes are as simple as

changing the pin numbers for connection to a device, such as connections between the Propeller and micro SD card slot are different on different development boards. Often, however, the code from several different objects would need to be merged in a way to get the desired result. For example, a module designed to use the Propeller Board of Education to record WAV files may use the on-board microphone, send it to the Propeller and use sigma-delta analog to digital conversion, and write that to an SD card. Though much of the code could be reused, it would need to be modified to use a different input source and different type of ADC.

Most modules built in Spin have one particular purpose. Keeping them single focused allows many modules to be used together. There is always a main module that is then used to call methods, Parallax's term for a sub-procedure. These methods can be in the main module or in other modules. The diagram of software module relationships is as follows:

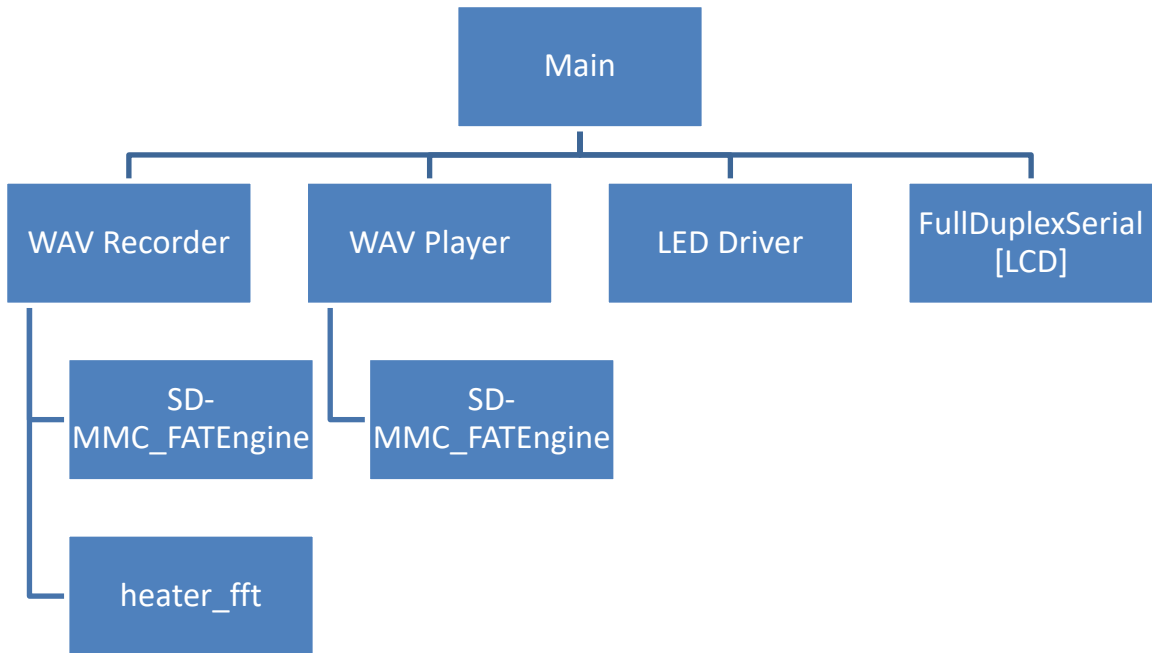


Figure 11 - Software Module Block Diagram

The main module sends parameters, including pointers to memory locations, that the other modules can then act on. The methods are run sequentially when called by the main

module unless a new cog is created to run them. If a new cog is initialized that then calls the methods, the main module continues executing while the new cog and its methods run separately in parallel. The Main and LED Driver modules were created for this project. FullDuplexSerial is a module supplied by Parallax with their compiler. The remaining modules were acquired from the Parallax Object Exchange. WAV Recorder and WAV Player were heavily modified, however; for instance, the original code used a sigma-delta analog to digital conversion and was replaced with code for the dedicated ADC chip, among other changes. Heater_fft was not functional when first downloaded so errors had to be corrected. SD-MMC_FATEngine was used without modification except for the removal of unused methods in order to reduce module size.

Implementation

Based on the design considerations, the proposed device should be able to be handheld. Finding a Propeller board that incorporates some, if not all of the desired component would simplify the design. The smallest board is the Propeller Mini. Though containing the Propeller, a crystal, program storage, and voltage regulators, all connections must be soldered. A device called the Project Board is a Mini with a USB connector and an area with through holes and pads to allow for prototyping. Next is the Quickstart. In addition to the components of the Mini, it has a USB connection for programming and power, 8 programmable LEDs, and 8 resistive touch buttons. All IO pins are available via sockets. An optional Human Interface Board can connect to the Quickstart. It contains a micro SD card slot, two PS/2 ports, an infrared transceiver, and multiple audio and video output connectors. Though it has some additional components that may be helpful, it has many that are unneeded.

The Propeller ASC+ contains a micro SD slot, a 12-bit ADC, and an external DC supply power jack. This board contains many of the additional components desired with the same footprint of the Quickstart, so it is the board that was selected as the basis for this device. There

are additional Propeller boards, such as the Board of Education and the Activity Board WX, but they are larger, more expensive, contain additional features not required, and are not as easily arranged and accessible for usage inside of an enclosure.

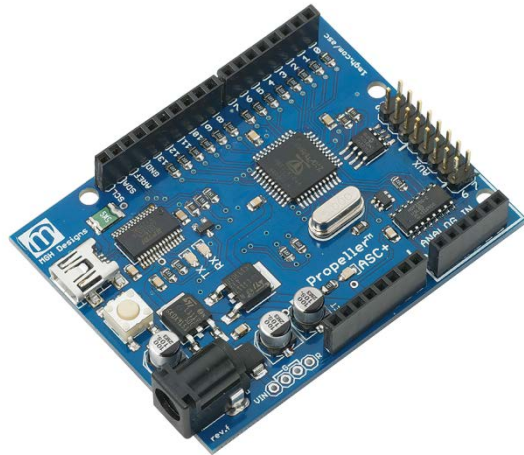


Figure 12 - Propeller ASC+ (21)

The geophone described in the previous chapter was chosen for further experimentation. A low frequency vibration is easier to generate than a low-frequency sound, so it will be easier to test the function of the device using the geophone as an input.

The design for the amplifier is modified from the MPLAB Starter Kit (Microchip Technology, Chandler, AZ) (18). Several changes were made to the design in order for it to meet the needs of this device. First, the MPLAB circuit had circuitry at the input that could provide power for a condenser microphone, if that was going to be used. That circuitry was removed. Secondly, the input signal was biased at 2.5VDC before going to the op amp. The same model op amp from the MPLAB circuit was used, an MCP6024, a Rail-to-Rail Input/Output, 10 MHz Op Amp. The same resistor values were selected, and the potentiometer is the same as the MPLAB circuit at 500kohms, which provides a gain from 2 to 229 V/V.

The anti-aliasing filter was designed using FilterLab 2.0 software from Microchip Technology (22). An anti-aliasing filter wizard is included in the tool that will design an

operational amplifier filter given certain parameters. The tool asks the user for the cut-off frequency, the sampling frequency, the resolution in bits of the analog-to-digital converter being used, and the desired signal-to-noise ratio of the final signal. On a final summary page, the software shows the parameters, and gives the user a choice between Butterworth and Chebyshev, with the filter order and the stopband attenuation listed. Once selected, the tool presents a schematic for the designed filter and a frequency response graphic for the magnitude and phase.

Infrasound is sound less than 30 Hz, therefore no signals greater in frequency than that are desired to be recorded or detected. For the filter tool, a cut-off frequency of 40 Hz was selected. Using 40 Hz instead of 30 Hz helped to keep the magnitude at 30 Hz the same as lower frequencies, i.e. no -3dB drop at 30 Hz.

With desired signals less than 40 Hz, the sampling frequency could be extremely low, e.g. 80 Hz using the Nyquist theorem. However, the design of a filter to reduce the magnitude far enough to prevent aliasing at 80Hz, yet maintain unity magnitude at 30 Hz, is extremely difficult. This software tool could not do it, stating it would be a filter order of greater than 8, which was out of the bounds of this tool to generate. It also would be hardware intensive, requiring much more space and power than desired. Researchers in (5) and (6) utilized sampling frequencies of 1 to 2 kilohertz in their infrasound research. This higher sampling frequency allows for a lower-order anti-aliasing filter to be designed because the magnitude of the output does not have to meet the stopband threshold until 500 Hz for a 1000 Hz sampling frequency. The researchers were also interested in capturing multiple harmonics of the infrasound, resulting in a passband of at least 150 Hz. This would also drive an increase in sampling frequency. Though 40 Hz was the selected cutoff frequency for this filter, a 1 kHz sampling frequency was selected to allow for a low order filter to be designed. The Propeller ASC+ ADC is 12 bits, so there was no ability to change that parameter in the filter tool. Setting the signal-to-noise ratio parameter to -65 dB created a filter order of 3, which only needed two op amps and 3 each of resistors and capacitors.

A Butterworth filter was selected because of the smooth frequency response in the passband.

The tool then output the schematic and frequency response graphics for this filter.

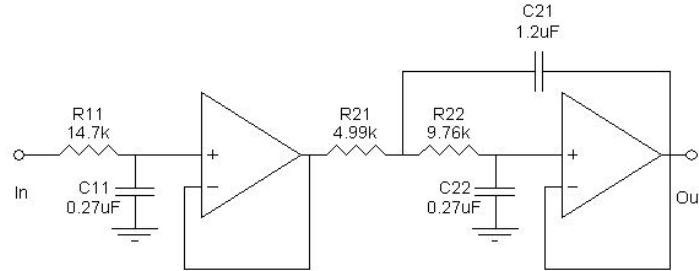


Figure 13 - Schematic of Anti-aliasing filter using FilterLab

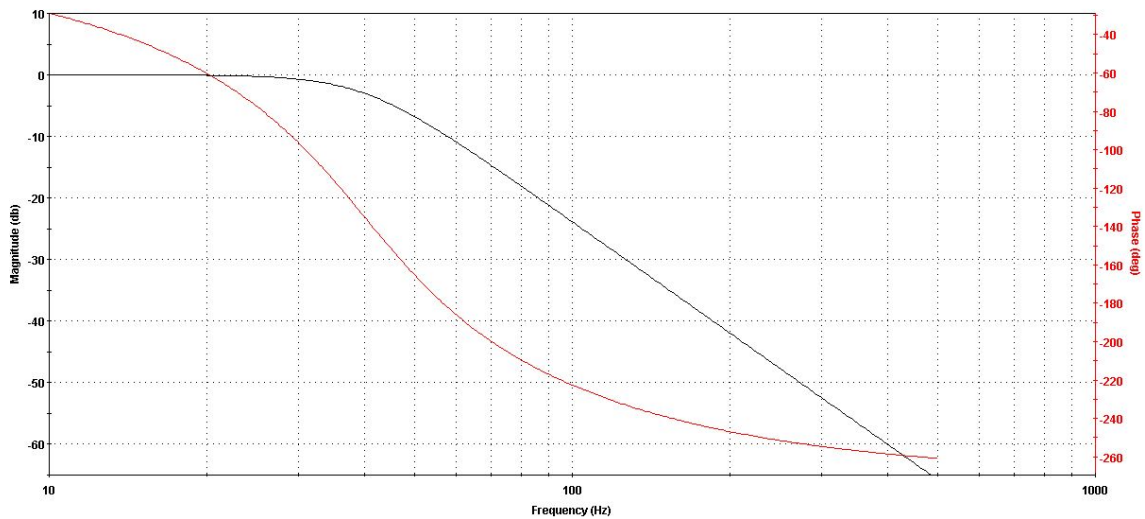


Figure 14 - Frequency Response of Anti-aliasing filter using FilterLab

The format for data storage on the SD card needed to be determined. Data could be written in human-readable form, such as ASCII characters. This could be the integer values that came directly from the ADC, or the calculated voltage levels based off of the ADC reference voltage. These values could be comma-separated values in a spreadsheet-style file stored to the SD card. The other option would be to store the raw data in a binary file. This could be accomplished by just writing the data as 16-bit words, or utilizing some other data format, such as a WAVE file. The human-readable versions would take more processing to turn the raw data into ASCII characters; therefore, it may not be able to be done at the speeds desired.

To test the speed of the ASCII-character writing, a program was written in Spin to time a write sequence. A new file was opened on the SD card. The current system counter value, CNT, was then stored to memory. One-hundred values were written to the SD card in ASCII integers, each separated by a new line, and the system counter was captured again. The average time to complete a write for a single value using this method was 148,241 clocks, or less than 539 samples per second. This human-readable value method of storing data is not compatible with recording audio data. The process of turning the raw ADC value into ASCII characters is much too time consuming.

The same experiment was performed, this time writing the raw data directly to the SD card. Using this method, the average number of clocks needed per sample was 435 clocks, which correlates to over 183,000 samples per second. A raw data format is the needed storage method.

The WAVE file format, a Microsoft and IBM audio file standard, is a fairly simple, uncompressed format for audio data (20). The file header contains information such as the sample rate, number of bits per sample, number of channels, and other information needed to properly describe the data in the file. Starting at byte 44, the raw data is written through until the end of the file. This file format will be the easiest to work with without requiring additional processing.

Before continuing with a WAVE recording program, it is important to understand how the Propeller's timing and cogs work. The Propeller has a 32-bit system counter (CNT) that increments every system clock cycle. All cogs have access to this value via the CNT register. Functions that need to be synchronized on different cogs can sync off of the CNT value. There is no need to have a cog set aside that only performs timing for the rest of the cogs. The Spin and assembly languages both have a command called `WAITCNT(cnt value)` which halts the processor until CNT equals *cnt value*. The cog will sit idle, waiting for the system counter to reach that

value. Once it does, execution will resume. Command usage such as `WAITCNT(50000 + cnt)` would read the current `CNT` value, add 50000 to that number, and then wait 50000 cycles before continuing (17).

The second way to use `WAITCNT`, and more valuable to this project, is for synchronized delay timing, such as what would be needed to accurately capture audio samples at evenly spaced intervals. The command looks more like `WAITCNT(Time += 50000)`. This usage requires the variable `Time` to have been set previously. The cog waits until `Time`, but then continues execution and adds 50000 to the value of `Time`. That command would then be followed by statements, such as capturing the ADC value. It would then be looped and return to the `WAITCNT` command, where it waits for the next 50000 clock interval. This form evenly spaces each `WAITCNT` by exactly 50000 clocks, therefore keeping perfect time between samples. Care should be taken to ensure that the statements following the `WAITCNT` command do not exceed 50000 clocks themselves. If they should, then the system counter would have passed the value of `Time`, and the `WAITCNT` procedure would halt all execution on this cog until the system counter loops and returns to the value of `Time` in the `WAITCNT` command. With a 32-bit counter, even at the Propeller maximum speed of 80MHz, it would take 53 seconds for the counter to loop back around to the “missed” value.

Cog function is another important aspect of the function of the Propeller. The Propeller does not have an interrupt function. Each cog on a Propeller is an independent processor core, complete with its own registers and RAM. Because of this multi-core design, there is no need to stop execution of the main program in order to execute an interrupt routine. What would be the interrupt routine can just be run on another cog, leaving the main program to continue execution without interruption. This ensures more deterministic timing, as there is no stopping and starting of the main program for interrupt events. There are several ways for this “interrupt routine” to happen at the needed time. As mentioned previously, the `WAITCNT` command will wait until a

specific system counter value is reached before continuing execution which works well for synchronized delay events. `WAITPEQ` halts execution until an input IO pin or set of pins matches an expected state. Execution of the rest of the code then commences. This function, as well as its inverse (`WAITPNE`), work well for halting execution until input values change, such as buttons being pressed or asynchronous inputs beginning. The input pins are checked every clock cycle, so it is a very responsive command. There is also a command, `WAITVID`, for use with the Propeller's video generation hardware. Lastly, a more brute force method is to read memory locations in Main RAM and wait for a certain condition to be met. A variable may get set in the Main program/cog, and a cog that is monitoring that variable's memory location could begin execution of code if that variable state changes. This is not quite as responsive as the previous commands, however, because of the way cogs get memory access. The Hub of the Propeller controls access to mutually exclusive resources, such as the Main RAM, to ensure system integrity. The hub runs at half of the system clock rate and gives access to each cog in a round robin fashion; therefore, there are 16 clock cycles between RAM access opportunities by a single cog. There could be up to 15 clocks of waiting before a cog can get access to the RAM if it missed the request window by 1 clock. Hub instructions also take more time to execute (8 cycles) so a RAM access takes between 8 and 23 clocks.

In order to accomplish high-bandwidth operations, a separate cog is started to perform that function. Cog startup, however, takes a bit of time itself. When a new cog is started, instructions and data from Main RAM are copied to the cog RAM. If assembly language is either the starter or what is started, it takes less than 9000 clocks, or a little less than 100us at 80 MHz. If Spin is launching a Spin cog, this can take about 25,000 clocks or 300 us (23). In either case, it is best to have a cog started and waiting for a trigger rather than try and start a cog the instant you need it to perform a function. This time delay could cause a `WAITCNT` to miss its mark and need another 53 seconds to wrap the system counter around again. This issue caused many early

problems when developing the audio sampling procedures. It could work correctly at sampling rates less than 3000 Hz, but when using higher sampling rates, it took many minutes to get very small blocks of data sampled. 3000 Hz is 333us, which is very near the cog startup time of 300us. Increasing the sampling frequency much more than that caused the WAITCNT value to be passed by the time the cog was started, so the system counter had to loop around.

The proper way to handle cogs is to start a cog before you need it, and have it wait for a trigger of some sort – either an IO value changing or even a variable set by another cog, whatever is practical for the function of the cog. Other cogs can run continuously, if needed. For instance, an object available for the ADC from the Parallax Object Exchange starts a cog in which the ADC runs continuously, taking samples and writing to a single Main RAM address as quickly as the cog can process it. Other cogs that want the current ADC value simply read from that memory address.

Now that the SD card data format and cog function is understood, the software can be written in order to access the SD card files. Several objects are available at the Object Exchange for writing to SD cards. Each has different capabilities and different levels of error-checking and file system handling, but the basics of reading and writing data are all the same. These objects typically have a start method which is called in order to create a new cog devoted to file handling. This cog can be started as the device is powered up and sit ready, waiting for calls to its methods to write and read data to the SD card. One of these objects called SD-MMC_FATEngine.spin.

Before beginning work on the WAV recorder, the accuracy of sampling needed to be tested; could the microcontroller capture a value at the exact same time interval every time? To test this, a program was written in Spin to loop 20 times. During each loop, the command `WAITCNT(Time += Period)` was called, and then the current system counter value was stored to a variable. The period was 10000 clocks, or 8000 Hz with the 80 MHz clock. After the loop

completed, the difference between sample values was calculated and the results sent to the terminal display on the PC. The difference between two values was 10000 clocks for all 19 intervals. Therefore, the `WAITCNT` command does exactly as claimed, and accurately spaced each sample.

Of the many objects available in the Object Exchange, there was a WAV recorder. However, this recorder used a sigma-delta analog-to-digital conversion technique that did not require a separate chip. To use the 12-bit ADC chip, the code had to be changed significantly to use a different method of analog-to-digital conversion. However, the overall structure of the code remained the same. Similar to the SD card object, this object also starts a new cog to initialize some settings, such as the sample rate. It also starts the ADC sampling, which continuously samples at the specified rate and writes the values to a data block in Main RAM. However, no files are being written to the SD card at this time. Once a method is called to start recording, a call to the SD card object opens a new file on the SD card, writes the header to the file, and then starts writing the data to the file. The data block to be written is broken into two 512-byte segments. These 512 bytes are the same size as a complete sector on the SD card. Flags are set by the ADC cog to signal which memory segment is being written to at the time. Once a segment gets filled, the flag is toggled, and the first cog then calls the `WRITEDATA` command to copy the segment to the file on the SD card. This process continues until a signal is given to stop the recording. At that time, the final file size is determined, appropriate data is written to the file header, and the file is closed. The SD card cog and ADC cog, however, are still always running and ready to begin another recording at any time.

This program was tested using the output from the headphone jack of an iPhone connected to the ADC input. The sampling rate was set to 22,050 Hz, or half of CD quality. A song was played from the phone, recorded to the SD card, and moved to a PC to hear the result. The song played back correctly, though with some noise. This test was repeated later with the

amplifier circuit. There was headroom in the prior test because the iPhone output did not drive the ADC input to its limits. With the amplifier circuit, it was able to be amplified to the maximum unclipped setting prior to recording, resulting in a louder sound when played back.

Writing data to a display is one of the easier aspects of using the Propeller. This device needed to remain small, so a 2-line, 16-character display was chosen. It was also designed by Parallax, so the interface is just as simple to control. ASCII DEC characters 32-127 are supported, as well as music tones using its built-in piezoelectric speaker. Whole character strings can be sent with a single command. A simple serial interface object written by Parallax and available on the Object Exchange can be used to communicate with the LCD display.

Frequency Analysis

A key component of the device is the ability to determine the frequencies included in the input signal. This analysis must be performed in real time in order to produce a timely alarm for elephant handlers in the event of elephant infrasound communication. Performing a Discrete Fourier Transform (DFT) on sampled input data can turn the sampled data into frequency information. Additionally, the utilization of the Cooley-Tukey algorithm, which follows, increases the efficiency and speed of performing the operation.

The complex Fourier series

$$X[k] = \sum_{n=0}^{N-1} x[n] \cdot W_N^{kn} \quad W_N = e^{-j\frac{2\pi}{N}} \quad k = 0, 1, \dots, N - 1$$

would normally require N^2 complex multiplications and additions. The algorithm developed by Cooley and Tukey can reduce the computations to less than $2N \log_2 N$ operations without the need for additional memory locations (24).

A Radix-2 FFT using decimation-in-time can accomplish this. Splitting the Fourier series into two sequences of even and odd indices results in

$$X[k] = \sum_{n \text{ even}} x[n]W_N^{kn} + \sum_{n \text{ odd}} x[n]W_N^{kn}$$

$$X[k] = \sum_{m=0}^{(N/2)-1} x[2m]W_N^{2mk} + \sum_{m=0}^{(N/2)-1} x[2m+1]W_N^{k(2m+1)}$$

With the following substitutions made:

$$W_N^2 = W_{N/2}, \quad f_1[m] = x[2m], \quad f_2[m] = x[2m+1]$$

The series can be rewritten

$$X[k] = \sum_{m=0}^{N/2-1} f_1[m]W_{N/2}^{km} + W_N^k \sum_{m=0}^{N/2-1} f_2[m]W_{N/2}^{km}$$

$$= F_1[k] + W_N^k F_2[k], \quad k = 0, 1, \dots, N-1$$

F_1 and F_2 are the $N/2$ DFTs of $f_1[k]$ and $f_2[k]$. Because of the periodicity of the summations and the symmetry of W_N , the series can be written

$$X[k] = F_1[k] + W_N^k F_2[k], \quad k = 0, 1, \dots, N/2-1$$

$$X[k + \frac{N}{2}] = F_1[k] - W_N^k F_2[k], \quad k = 0, \dots, \frac{N}{2}-1$$

This rearrangement of the equation cut the number of multiplies needed in half (25 p. 457). The recursive nature of this method then allows the $N/2$ DFTs to be decimated themselves,

repeating the process until there is only a 2-point DFT to compute. Most importantly, this allows computations to occur in-place without the need for additional memory. Once the calculations are made there is no need for the original input pair. These computations are called butterflies due to their diagram. Given complex input pair (a,b), the calculation is

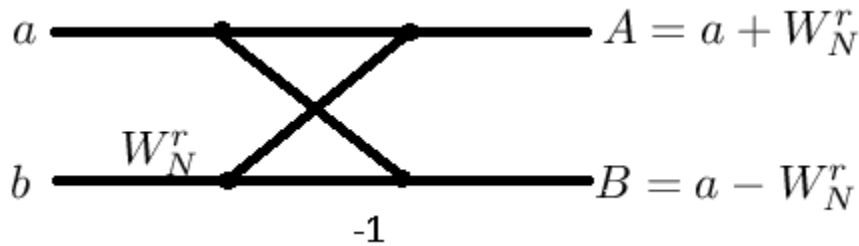


Figure 15 - Butterfly diagram of a Fast Fourier Transform

This Fast Fourier Transform (FFT) method has been created in a module available on the Parallax forums (<https://forums.parallax.com/discussion/128292/heater-s-fast-fourier-transform>.) The module uses the Cooley-Tukey method to perform the transform. It is written to take a 1024 sample input. It is capable of using a different sample size; however, changes would have to be made to the twiddle factors. To use the module, the addresses to two long-sized buffers are sent: **bx**, which contains the samples to transform, and **by**, used to perform the complex calculations and must be initialized to zeros. The steps to perform are also sent to the module via pre-defined command bits. These commands include CMD_DECIMATE, CMD_BUTTERFLY, and CMD_MAGNITUDE. All operations of the module are executed using assembly code, which greatly increases the speed of computation compared to using Spin.

CMD_DECIMATE performs a radix-2 decimation in time, reversing the order of the bits as per the Cooley-Tukey method. CMD_BUTTERFLY then performs the FFT through the different stages. CMD_MAGNITUDE converts the complex output to a magnitude number stored in **bx**.

The first 512 indices of **bx** contain the magnitudes of each frequency bin. If the sampling rate is equal to the number of samples in the FFT, then the indices in **bx** equal the frequency in hertz of the magnitude results. For example, if the sampling rate is 1024 Hz, and a 1024-sample FFT is performed, then the magnitude found at index 30 of **bx** is the magnitude at 30 Hz. No additional calculations would need to be made to convert the frequency bins to a range of frequencies in Hz. To ensure the FFT performs as desired, a software-derived waveform of known frequency was loaded in the **bx** array. A 16 Hz waveform ranging in magnitude from -2047 to +2047 was input and the resultant magnitude in index 16 was 2046, with magnitudes of 0 for all other indices.

Device Assembly

The Propeller ASC+ card is a rather small board at 2.70 x 2.10 in (68.6 x 53.3 mm). It is Arduino-shield compatible; therefore, it has headers for all IO pins as well as for VIN to the board's regulators, ground, and +5 VDC out. A perfboard with compatible headers was connected to the ASC+. This board, from here forward called the main board, contains the amplifier, anti-aliasing filter, DAC, and LED driver chips, as well as connector jacks for cables to other boards. This board has approximately the same dimensions as the ASC+.



Figure 16 - Infrasonic Detection Device

A plastic enclosure of size 4.724" L x 3.157" W (120.00mm x 80.19mm) X 2.311" (58.70mm) contains all components. Slots and holes were cut into the enclosure to allow for connections. The connected boards were mounted in the enclosure to allow access to the USB port, the SD Card slot and DC power connector. The four control buttons were placed on perfboard and connected to the main board via a cable and affixed to the enclosure cover. Likewise, the LED bar was put on perfboard, affixed to the cover and connected to the main board via a cable. The LCD Display was also affixed to the cover and connected to the main board via cable. The potentiometer, input jack, and output jack were wired directly to the main board and attached to the enclosure. An AAx4 battery holder was attached to the inside of the enclosure. It was connected to an external power switch and its wires connected via jumper-style pin to the VIN and GND headers on the main board.

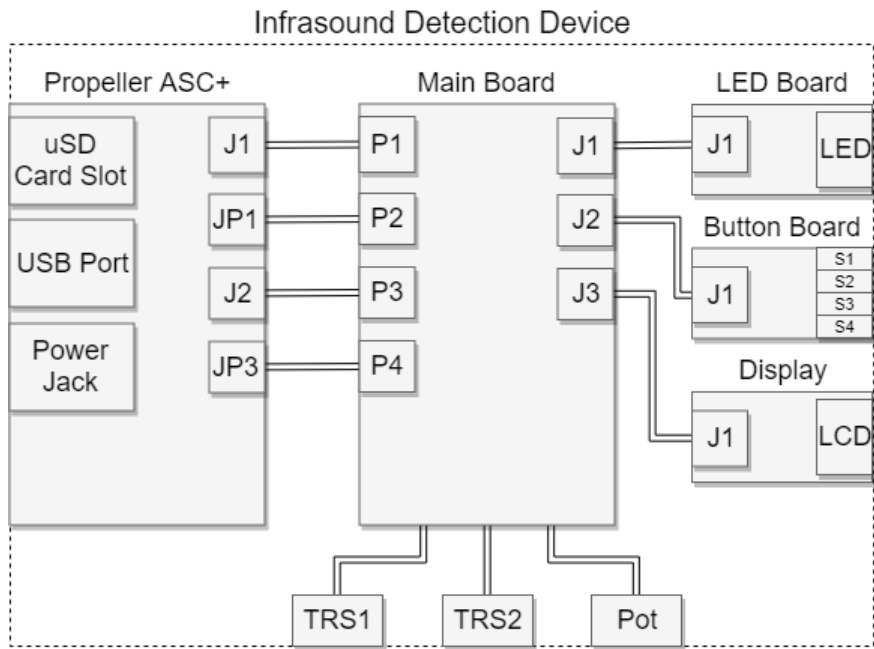


Figure 17 - Block Diagram

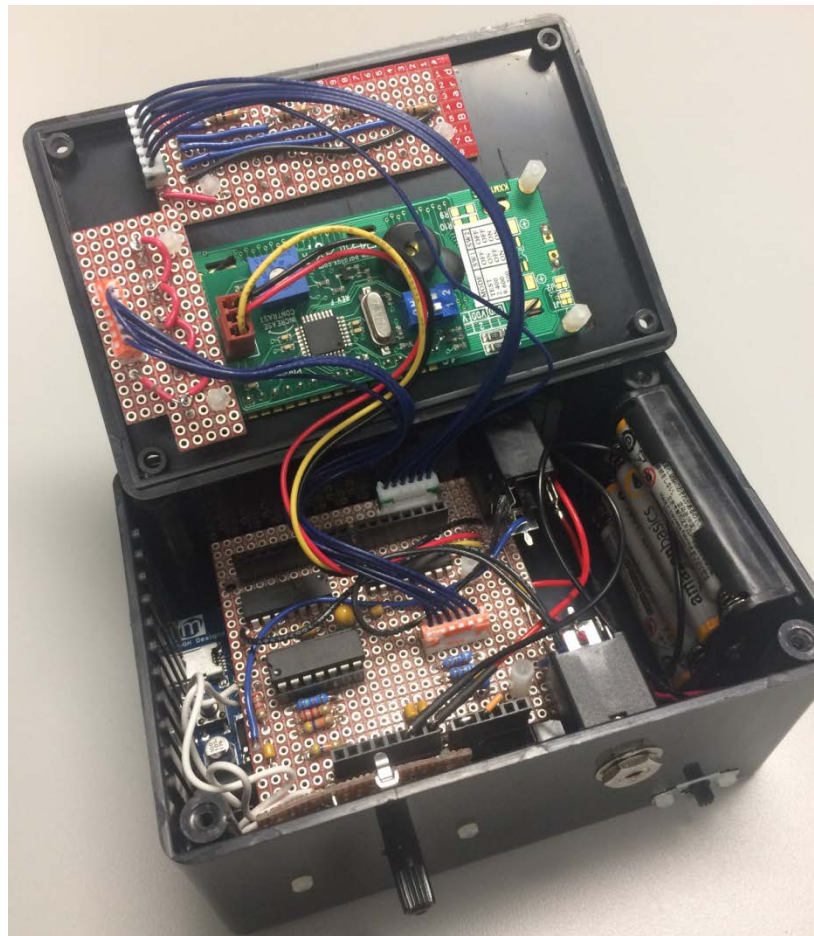


Figure 18 - Infrasound Detection Device (Cover Open)

Device Specifications

Power – 6-9 VDC external power jack; OR mini USB jack; OR 4 x AAA

Input – 3.5mm audio jack (mono); 0-1.25 peak-to-peak VAC; 2Hz – 10,000 Hz

Output – 3.5mm audio jack (mono)

Storage – micro SD card

Physical Dimensions: 4.7 in x 3.5 in x 2.5 in.

Weight: 10.5 ounces

CHAPTER IV

RESULTS

In this chapter the device is tested to determine if it meets the objectives. The first objective to test is the ability to accurately record infrasound. The second objective to test is the ability as a learning device and/or for field study. Last, the objective to accurately monitor infrasound frequencies will be tested.

The results showed that the device was able to record infrasound for extended periods of time. The device is easy to use and could be used for field usage, though the current battery capacity is limited. Testing also showed that infrasound was accurately detected and triggered the alarm.

Recording

The recording capabilities of the device were tested using several input devices and two modes. The recording modes of the device are 1) Record continuous, until stopped or maximum file length is reached, or 2) Timed recording, where the recording stops after the selected timer is complete.

The maximum file length in number of bytes is 2^{31} bytes, or 2 GB. This is the maximum positive number that can be held in a long memory location, therefore the largest number the Propeller can track. The sampling rate on the device was set to the highest value (20,000 samples per second) and a continuous recording started. The file stopped with size 2,097,152KB, which is 2 GB. The length of the file was 14 hours, 54 minutes and 47 seconds. This test was conducted utilizing external power through the USB port.

The second test for recording was to examine the length of time the batteries could sustain recording. The device uses 4 AAA batteries providing 6 VDC. A regulator onboard the Propeller ASC+ board provides 5 VDC for the rest of the device. To conduct this test, the sampling rate was set to 1000 Hz and new Amazon Basics AAA batteries were installed. The record continuously option was started and the recording occurred until the batteries died. The resulting file stored on the SD Card was incomplete, since it could not be closed properly by the file system. However, based on the number of bytes written, it could be determined how long the device recorded before power down. This experiment was conducted three times.

| Test | File Size (KB) | Seconds | Duration |
|------|----------------|---------|----------|
| 1 | 37280 | 19087 | 5:18 |
| 2 | 37952 | 19431 | 5:23 |
| 3 | 42464 | 21742 | 6:02 |

Table 1 - Battery Duration Test

The results were not as good as expected. After just two hours, the display had faded completely. A DMM was connected to the device in order to monitor the current being drawn during operation. After initial startup, the device drew 58 mA from the batteries. During recording or playback operations, 80 mA were drawn. The current draw would spike to over 200 mA when the piezospeaker was being used.

The capacity of a battery in milliamp-hours (mAh) depends on the current it is discharging. An Energizer Max AAA has a capacity of 1100 mAh at a continuous current of 25

mA if dropping its output voltage to 0.8 VDC from the original 1.5 VDC. However, at 100 mA, it only has a capacity of just over 900 mAh (26). The same specification sheet contains a graph that shows the typical usage for a digital audio device at 50 mAh and the voltage vs. hours curve for that usage. The curve shows that the voltage will drop to less than 1.25 VDC in about 10 hours of use. Considering that the bare minimum usage by the device when the display is the only function executing is 58 mA, it follows that it will take even less time to drop to 1.25 VDC, which with 4 batteries gives 5 VDC, the bare minimum needed to drive the 5-volt components.

The third recording test was to judge the accuracy of the timed recording. A record time of one minute was selected. Two recordings were made at 1000 Hz and two recordings at 20,000 Hz. Once complete, the WAV files were opened on a PC using Audacity audio software. The software reported the recording length to be 1 min 0.416 sec for the 1000 Hz files and 59.507 and 59.149 sec for the 20,000 Hz. The variability can be explained because after the timer is started, the recording cog must complete a 256-sample block before it starts sending the data to the SD card. At a sampling rate of 1000 Hz, it would be over a quarter second of delay before recording. Likewise, at the end of the timer, recording is only shut down between 256-sample blocks. This could potentially add another quarter second of recording time.

The final recording test was for recording quality. The low pass filter on the device can be bypassed using a jumper. This allowed recordings in normal human aural range to be assessed for quality. The sampling frequency was set to 20,000 Hz and songs were played from an iPhone headphone jack through a cable to the input jack on the infrasound device. The gain knob was adjusted to get the amplitude in the proper range according to the LED readings. The songs were recorded and the SD card then taken to a PC to playback using its speakers. The songs were faithful representations of the original input from the iPhone. The low pass filter was included back into the signal back and recordings of signals from a function generator were made.

A 22 Hz signal from a function generator was recorded. The resulting WAV file was analyzed on a PC using Audacity software. The signal was a very clean sine wave with an exact frequency of 22Hz with no variation. The device is able to record very low frequencies accurately.

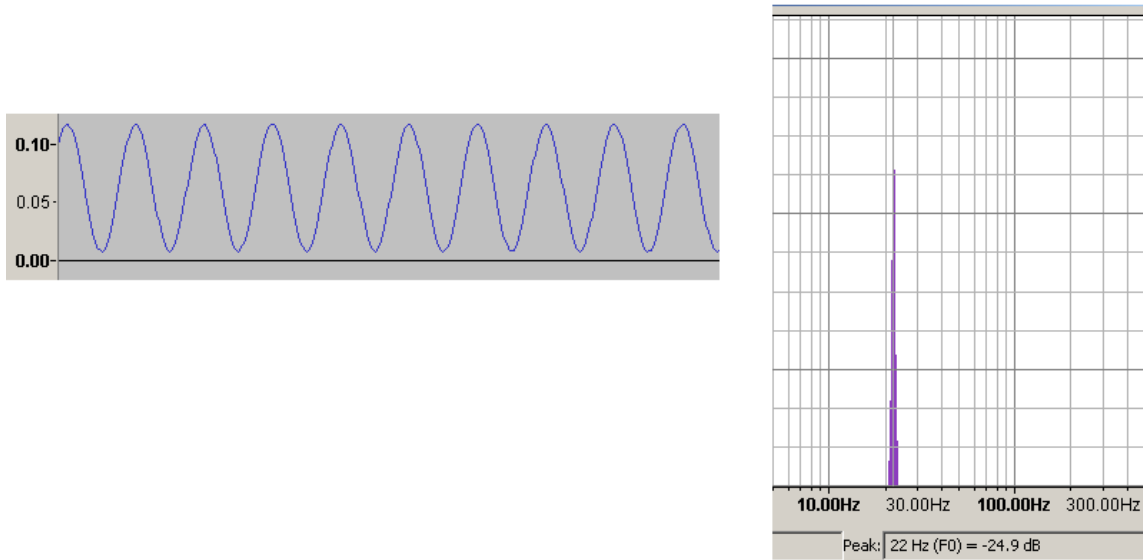


Figure 19 - Signal and Frequency Spectrum of Recorded WAV File (22 Hz Signal Input)

Playback

The second objective to be tested was usability in the field. The recording portion of this objective has been shown. However, feedback is needed by the user to ensure that the recordings are in fact being made and sound proper, e.g. volume loud enough, not over-driven, etc. The recordings made previously were used in order to test the playback function of the device. Earbuds were connected to the output headphone jack. The interface was used to select a file on the SD card and played. Music files played the song accurately; however, there was a noise included in the sound which was worse on the beats of the song. There seems to be an issue with how the signal is sent from the DAC to the headphones. The only circuitry from the output of the DAC is a 0.1 uF capacitor in series to block the DC output of the DAC, since the signal is biased at 2.5 VDC.

One of the WAV files was a 44 Hz signal. Though technically it is not infrasound, it is still inaudible using earbuds because they are unable to produce low frequencies well. Pressing the 10x button on the interface played the tone back, now at 440 Hz. This matched the musical tone of “A”. The pitch matched when compared against an “A” generated by a tuner; therefore, the speed up function is accurate and allows the listener to “hear” infrasound. This, however, shortens the length of the signal, so an elephant call that is less than 10 seconds is now less than 1 second in length; however, it does let the user know that a sound was captured, which is the intent of the increased speed function. These signals are also plagued with a noise that prevents the signal from being a clean sound; however, the recorded sound is discernible from the noise.

Infrasound Detection

The most important feature to test was the infrasound detection. Several input types were used, including pure tones, recorded elephant calls, and the geophone.

Pure Tones

An online tone generator (<http://www.szynalski.com/tone-generator/>) was used to generate sine wave signals for device input. The output of a PC was connected to the device input jack. The initial tone was set to 30 Hz, the gain adjusted, and then the frequency was changed and FFT results output to a serial terminal on the PC. The results of the experiment seem to show the weakness of the PC sound card in its ability to produce signals in the infrasound range. The frequencies detected were accurate, e.g. a 24 Hz tone was detected at 24 Hz by the device; however, the amplitudes of the signals varied greatly depending on the frequency generated, with a noticeable notch at 40 Hz.

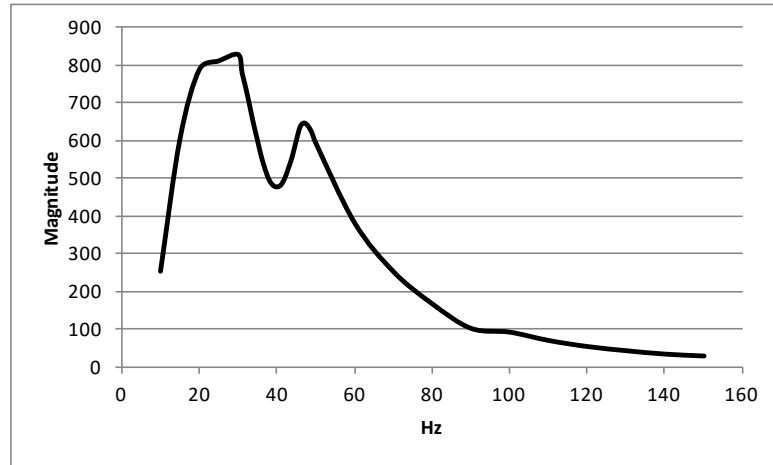


Figure 20 - Frequency Response of Device Using Sound Card Output

Due to the poor results of the PC tone generator, a function generator was used to provide a more accurate input. Although the notch at 40 Hz was no longer present, there was still significant signal loss starting at 40 Hz toward DC, very similar to the sound card source. It was found that the input coupling capacitor and impedance of the amplifier circuit were acting as a high-pass filter with a corner frequency of

$$f_c = \frac{1}{2\pi RC} = \frac{1}{2\pi(23440)(0.1\mu F)} = 67.93 \text{ Hz}$$

which was filtering out all of the infrasound, but especially the lowest frequencies. By changing the input coupling capacitor to 1.1 μF , the corner frequency was now 6.2 Hz, which allowed the desired frequencies to pass through.

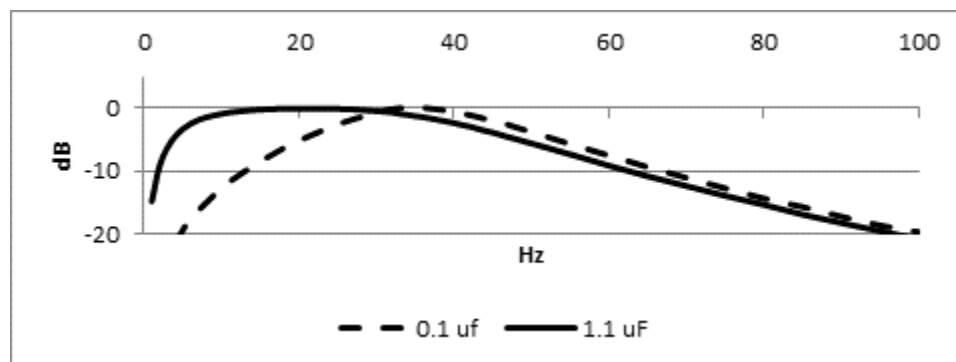


Figure 21 - Frequency Response Using Function Generator for Two Different Coupling Capacitor Inputs

Elephant Calls

Despite the amplitude issues of the PC sound card noted earlier, recordings of elephant calls played into the device would provide a more realistic signal to test infrasound detection capabilities. ElephantVoices, an elephant advocacy group, has a database of elephant calls on their website (27). Their collection includes recordings of 7 rumbles, which are the calls containing infrasound. The following table gives their descriptions.

| Name | Description | Length |
|----------|---|--------|
| A3203414 | A relatively long, undulating rumble by a juvenile female | 0:06 |
| B1400110 | A soft, short rumble by a calf | 0:05 |
| B3304802 | A long, powerful and highly modulated rumble by an adult female | 0:06 |
| C2312431 | A long, unmodulated rumble by an adult female | 0:09 |
| F1200221 | A throaty rumble with a roaring quality by an adult female | 0:05 |
| U1605722 | A long pulsating rumble by an adult male | 0:07 |
| U1700443 | A short breathy rumble by an adult male | 0:02 |

Table 2 - Elephant Calls

The gain was set with the peak of the input between 60-80% of maximum magnitude with A3203414. The MP3 of the elephant call was played on a PC and sent via the headphone jack to the input of the device. The device was put into monitor mode, and the results of the FFT sent to the serial terminal and then transferred to MATLAB and plotted.

The MP3 was then input into MATLAB where it was resampled at 1024 Hz and an FFT performed on the data, which was then also plotted. The following figure shows a few comparisons of the MATLAB FFT results (on the left) to the device's FFT results (on the right.)

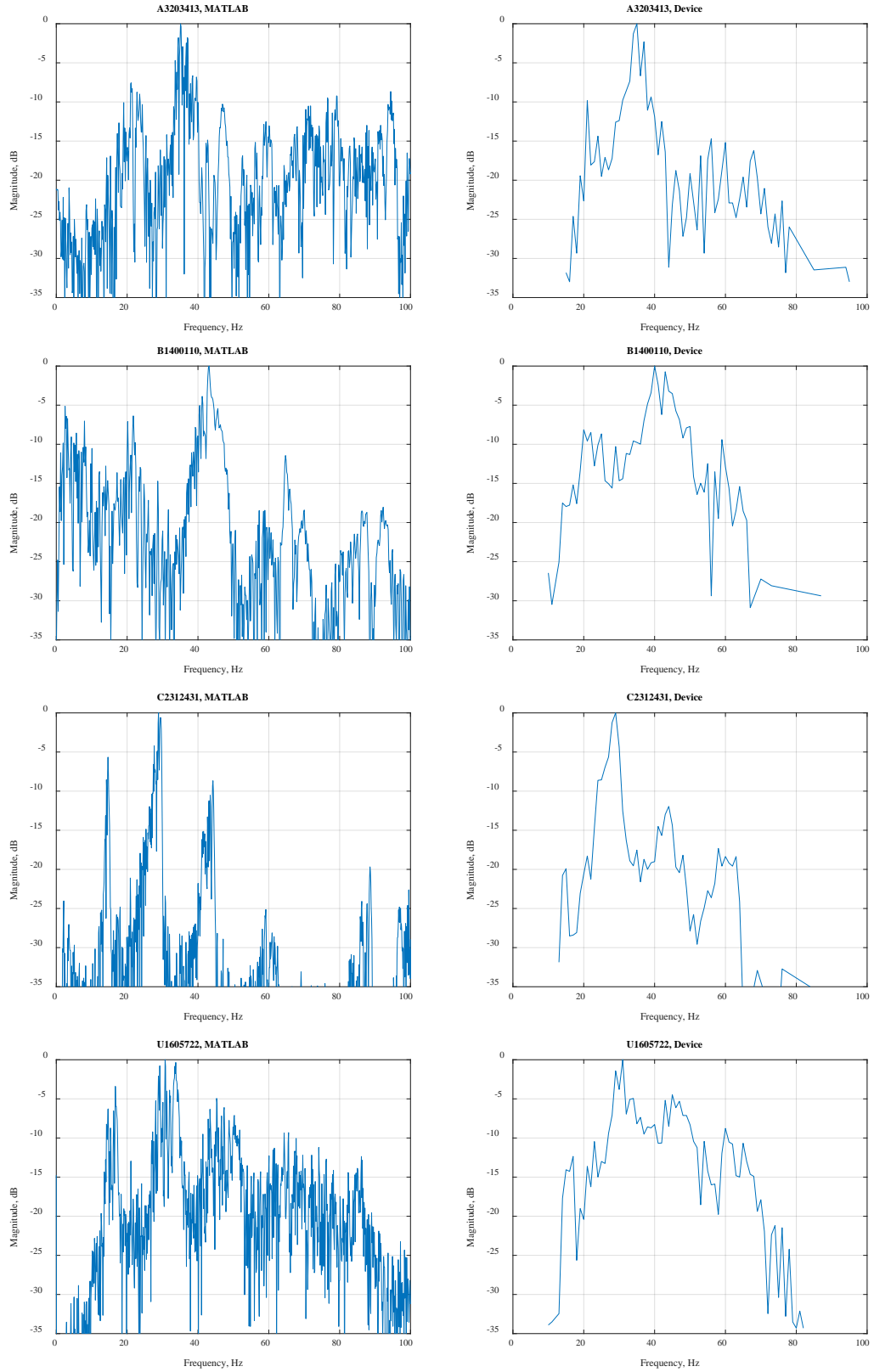


Figure 22 - Frequency Analysis of Elephant Calls

The MATLAB-generated FFT had more bins than the device FFT, which only had integer bins; therefore, the device FFT results look smoother. The FFT on the device accurately determined the frequencies that were prominent in the elephant call. For example, A3203413 had a peak at around 35 Hz which is clearly shown on both FFTs. The same can be shown on the other examples, as well.

When in the Monitor/Alarm mode, the device runs the FFT on 1 second of samples. If it detects a magnitude greater than the threshold (102 at a 10% detection threshold) at a frequency at or below 40 Hz, the alarm is tripped. It then displays the magnitude and frequency of the peak magnitude. The results of each elephant call were as follows:

| Name | Results | Detected? |
|----------|--|-----------|
| A3203414 | 203 at 37 Hz, 190 at 37 Hz, 458 at 36 Hz | Yes |
| B1400110 | 117 at 26 Hz, 103 at 34 Hz, 139 at 38 Hz | Yes |
| B3304802 | 104 at 36 Hz, 146 at 30 Hz, 147 at 31 Hz | Yes |
| C2312431 | 480 at 25 Hz, 610 at 29 Hz, 423 at 28 Hz | Yes |
| F1200221 | 106 at 36 Hz, 106 at 36 Hz, 109 at 36 Hz | Yes |
| U1605722 | 116 at 40 Hz, 131 at 37 Hz, 134 at 37 Hz | Yes |
| U1700443 | 138 at 36 Hz, 103 at 36 Hz, 108 at 35 Hz | Yes |

Table 3 - Infrasonic Monitoring Results

All 7 elephant calls were detected. The upper limit frequency is 40 Hz. Though that is not technically infrasound, low-frequencies are harder to discern. According to the ISO 226:2003 equal loudness curves, frequencies below 50 Hz would have to be over 35 dB louder than a normal conversation to have the same effective loudness. The higher-than-infrasound threshold was chosen to account for this.

Geophone Test

To test the geophone for infrasound detection capability, a rig was constructed to suspend the geophone to allow for free movement. A small vibration motor was attached to the geophone and powered to provide a slow turn rate, less than 1200 RPM, which simulates a vibration less

than 20 Hz. The infrasound-detection device was set to monitor mode. It triggered the alarm several times, displaying detected frequencies around 29 Hz each time.

CHAPTER V

CONCLUSION

In this work, it was shown that a low-cost device can be built which is capable of recording various analog inputs. The device can perform frequency analysis on those inputs and provide feedback to the user. The device is also shown to be low-cost and able to record for extended periods of time left unattended. The three objectives of this thesis – a low-cost infrasound warning device, an extended-period recording capability, and an educational tool – were met.

The total cost of components for the device was under \$120. \$75 were only for the Propeller board and the display; the remaining \$45 were for the remaining electronics, interface, and enclosure components. If this were a production item, a designer could develop their own board instead of utilizing the ASC+ prototyping board used during this project. That board could be much less expensive because only the desired components would be included.

Future Work

This device was only a prototype. More work could be done to enhance some features of the device. Using a printed circuit board to build the circuitry would enhance the look, clean operation, and reliability of the device. Additional circuitry may need to be added and/or software changed to enhance playback to remove the noise that was present in this device. A slightly larger enclosure to better accommodate the internal components would make the device easier to build and maintain. That larger enclosure may also allow room for larger batteries, would help in allowing the device to be powered internally for longer periods of time.

Just simply switching to AA batteries may increase powered-on time to over 15 hours. Likewise, an external battery source could be plugged in using the DC power jack for a much longer usability period. The device can take an input of up to 9 VDC. This would also allow the device to stay in place without the external battery pack needing swapped in and out.

The ability to identify and eliminate constant background noise, such as an air conditioner, would be helping in preventing false alarms when infrasound from those sources would be able to surpass the alarm threshold.

The most important future work would be on the input sensor. The experimentation with the geophone showed that it is able to detect vibration; however, that does not seem to be the most efficient sensor for infrasound, nor is it capable of recording the actually infrasound. Without a sensor that can consistently and accurately detect infrasound, the safety capability of this device would be in question. Finding a microphone or other input sensor for low-cost would be an important piece of the continuing work.

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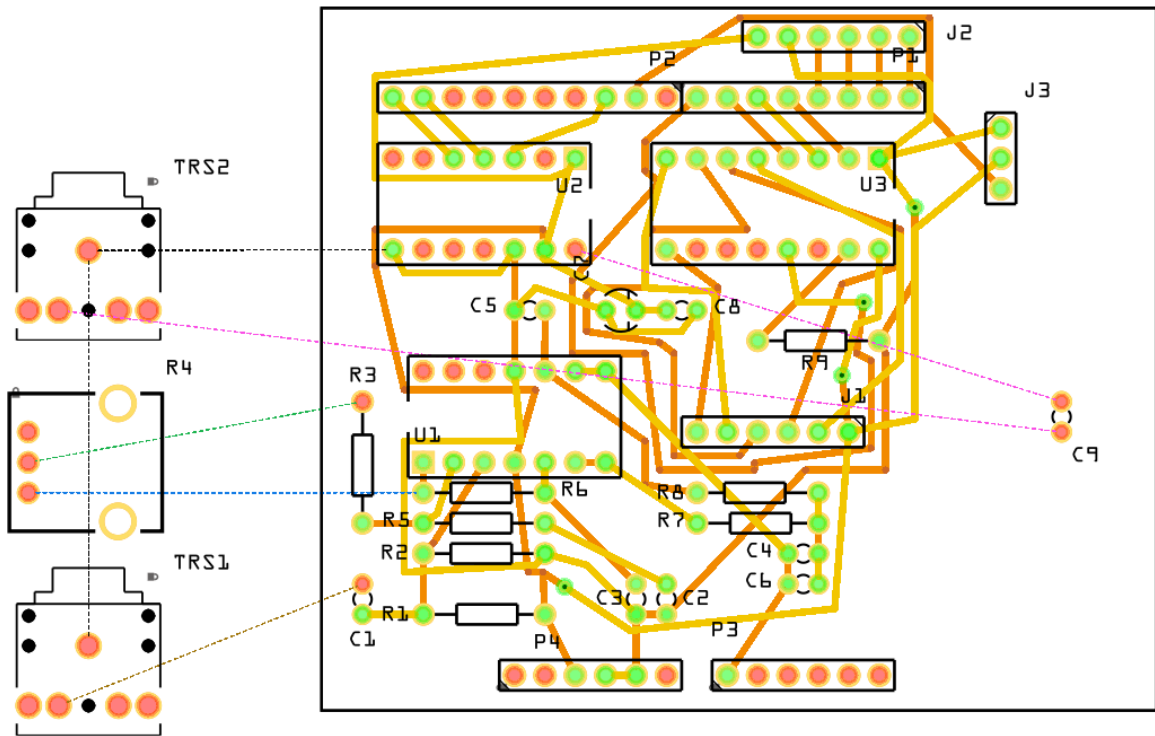
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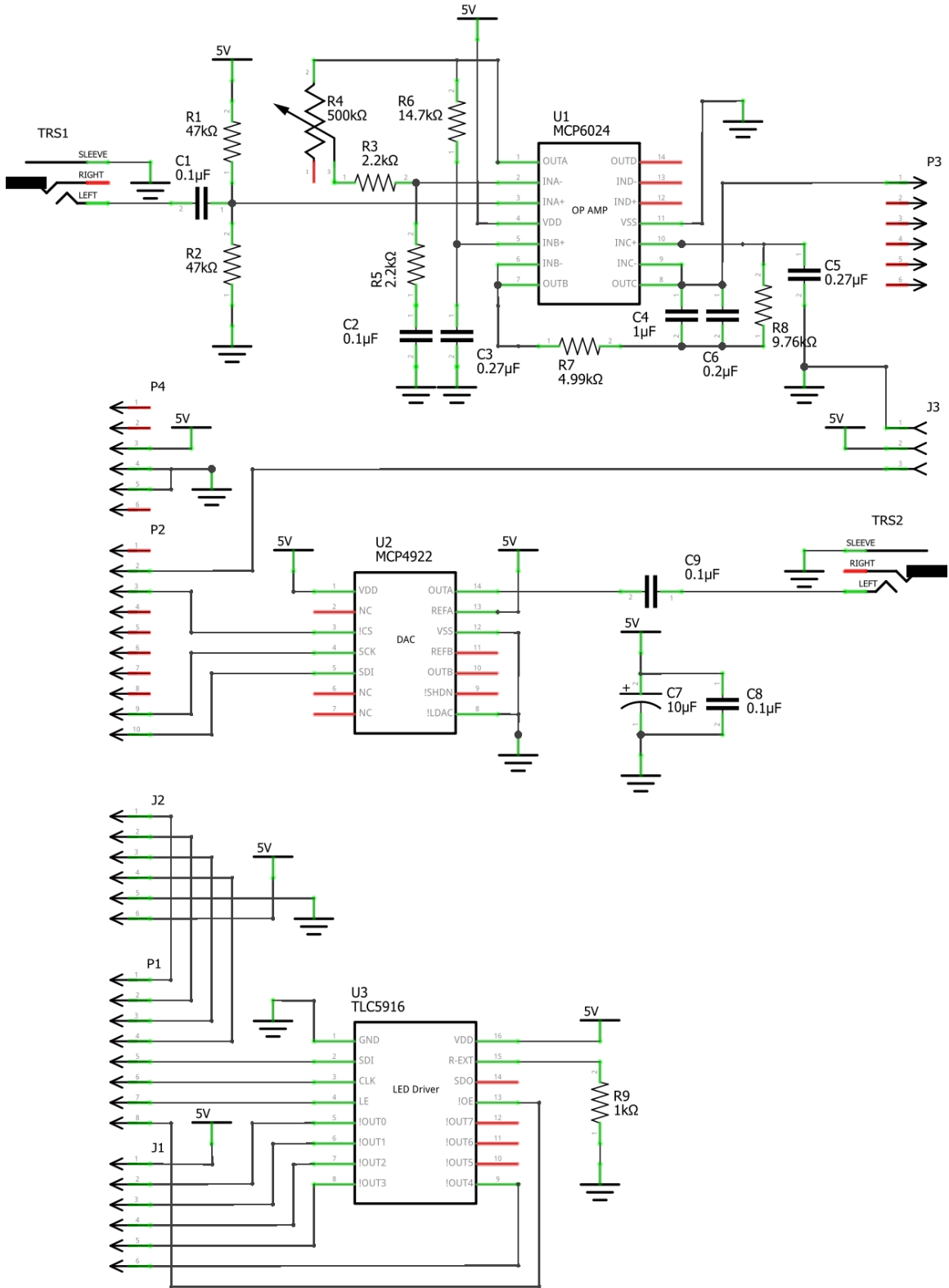
APPENDICES

A

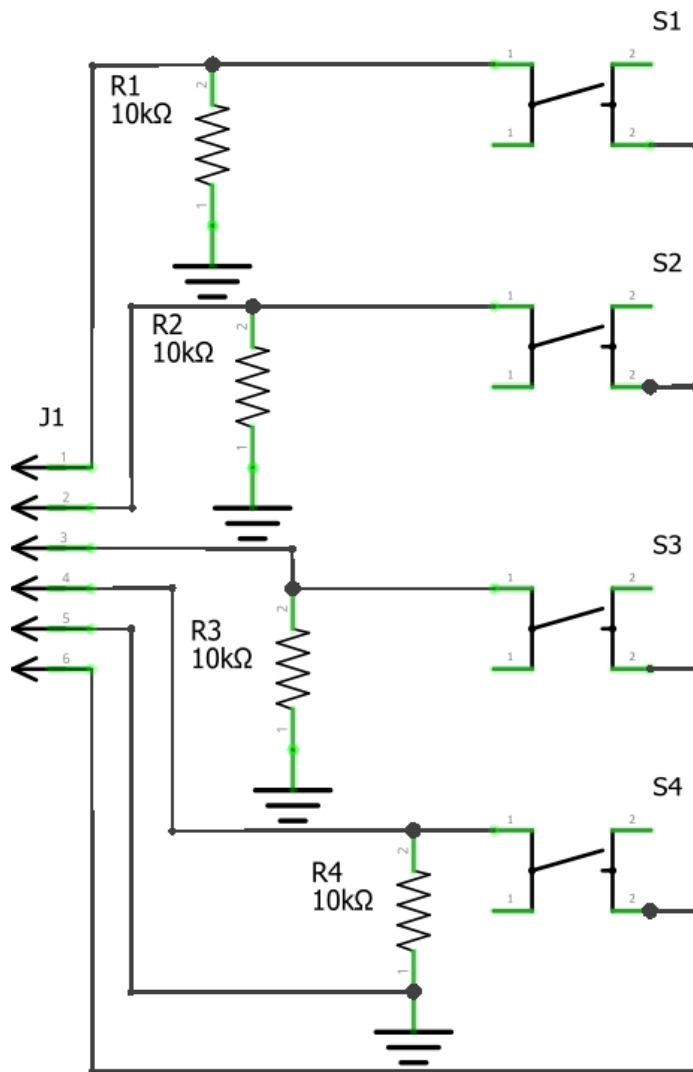
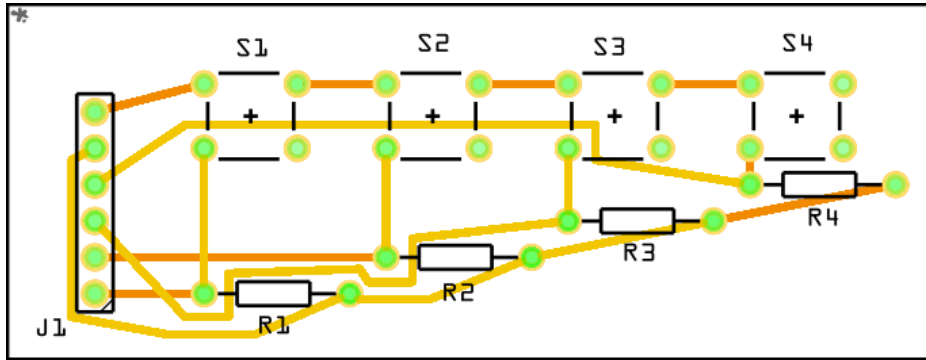
SCHEMATICS AND LAYOUT

Main Board

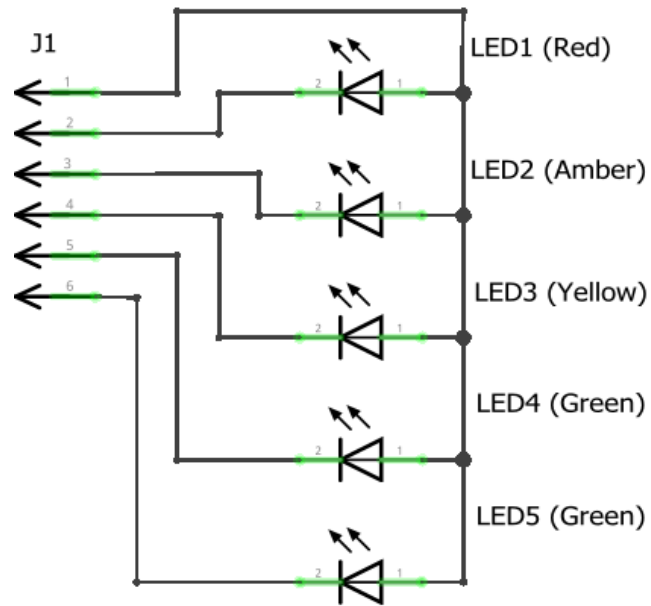
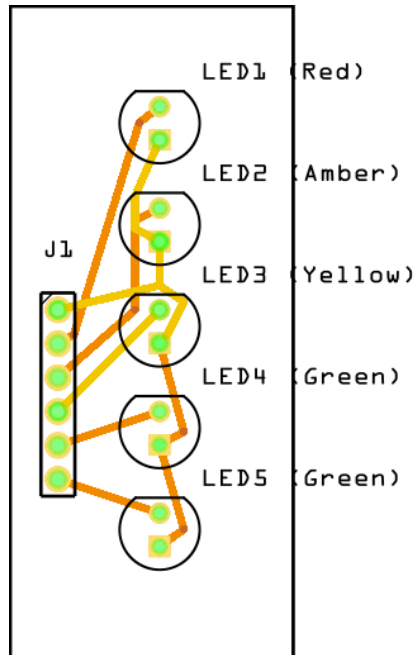




Buttons Board



LED Board



B

PARTS LIST

| PN | Manufacturer | Description | Qty | Price |
|--------------------|-------------------------------------|--|-----|----------|
| 32214 | Parallax | Propeller ASC+ board | 1 | \$ 49.99 |
| PRT-11417 | SparkFun Electronics | Protoboard Snappable | 1 | \$ 7.95 |
| PRT-13268 | SparkFun Electronics | Arduino Stackable Header Kit | 1 | \$ 1.50 |
| 1591TSBK | Hammond Manufacturing | BOX PLASTIC BLK 4.72"L X 3.16"W | 1 | \$ 6.50 |
| 27977 | Parallax | Parallax 2 x 16 Serial LCD (Backlit) | 1 | \$ 24.99 |
| A06KR06KR26E152A | JST Sales America | JUMPER 06KR-6S-P - 6" | 2 | \$ 2.56 |
| B6B-PH-K-S(LF)(SN) | JST Sales America | CONN HEADER PH TOP 6POS 2MM | 4 | \$ 1.32 |
| ACJS-MV35-3 | Amphenol | CONN JACK STEREO 3.5MM | 2 | \$ 1.92 |
| SSA-LXB525-G2YAID | Lumex | LED Bars and Arrays LED Bars and Arrays 1.8x5.3mm 5 Unit LED Green/Ylw/Amb/Red | 1 | \$ 3.58 |
| 1825910-7 | TE Connectivity ALCOSWITCH Switches | SWITCH TACTILE SPST-NO 0.05A 24V | 4 | \$ 0.40 |
| 2482 | Keystone Electronics | HOLDER BATTERY 4CELL AAA 6" LEAD | 1 | \$ 2.23 |
| TLC5916IN | Texas Instruments | IC LED DRIVER LINEAR 120MA 16DIP | 1 | \$ 1.37 |
| MCP6024-E/P | Microchip Technology | IC OPAMP GP 10MHZ RRO 14DIP | 1 | \$ 2.01 |
| MCP4922-E/P | Microchip Technology | IC DAC 12BIT DUAL W/SPI 14DIP | 1 | \$ 2.70 |
| 296UD504B1N | CTS Electrocomponents | POT 500K OHM 0.15W CARBON LINEAR | 1 | \$ 1.54 |
| CF14JT10K0 | Stackpole Electronics | RES 10K OHM 1/4W 5% AXIAL | 4 | \$ 0.40 |
| CF14JT47K0 | Stackpole Electronics | RES 47K OHM 1/4W 5% AXIAL | 2 | \$ 0.20 |
| MFR-25FBF52-14K7 | Yageo | RES 14.7K OHM 1/4W 1% AXIAL | 1 | \$ 0.10 |
| CF14JT2K20 | Stackpole Electronics | RES 2.2K OHM 1/4W 5% AXIAL | 2 | \$ 0.20 |
| MFR-25FBF52-4K99 | Yageo | RES 4.99K OHM 1/4W 1% AXIAL | 1 | \$ 0.10 |
| CF18JT1K00 | Stackpole Electronics | RES 1K OHM 1/8W 5% CF AXIAL | 1 | \$ 0.10 |
| MFR-25FBF52-9K76 | Yageo | RES 9.76K OHM 1/4W 1% AXIAL | 1 | \$ 0.10 |
| A14042900UX0338 | Uxcell | CAP TANTALUM 10U 35V RADIAL | 1 | \$ 0.67 |
| C440C105M5U5TA7200 | Kemet | CAP CER 1U 50V Z5U AXIAL | 2 | \$ 0.82 |
| SA115E274MAR | AVX | CAP CER 0.27U 50V Z5U AXIAL | 2 | \$ 2.10 |
| FG24X7R1H224KNT06 | TDK Corporation | CAP CER 0.22UF 50V X7R RADIAL | 1 | \$ 0.29 |
| SA115C104KARC | AVX | CAP CER 0.1U 50V X7R AXIAL | 3 | \$ 0.78 |

Total \$116.42

!Main.spin

```

{}
//////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
// This program is the top level module for the Infrasound detection device. It handles
// the LCD, LED, and button interface.
//////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
}}

CON
    _clkmode = xtall + pll16x      'Standard clock mode * crystal frequency = 80 MHz
    _xinfreq = 5_000_000

    _dopin = 12
    _clkpin = 13
    _dipin = 11
    _cspin = 8
    _cdpin = -1 ' -1 if unused.
    _wppin = -1 ' -1 if unused.

    _rtcres1 = -1 ' -1 always.
    _rtcres2 = -1 ' -1 always.
    _rtcres3 = -1 ' -1 always.

    _rate = 1000 ' Default sample rate.
    '_rate = 8000
    '_rate = 22050

    _thresh = 10 'Default Threshold level

VAR
    long i,j,k,SamplingFreq, AlarmThresh,FileName, maxfreq, maxamp, Stopped, CardCheck
    long AlarmOn, RecTime, PlayFlag
    long maxValue                    'maximum value of current sample block
    long Stack[100]
    long x10

OBJ
    LCD      : "FullDuplexSerial.spin"
    adc      : "WAV-Recorder.spin"
    LED      : "LED Driver.spin"
    dac      : "WAV-Player.spin"

PUB Main
    'LCD.start(9, 9, %1000, 19_200)
    LCD.start(9, 9, %1000, 9_600)      'initiate LCD screen communication
    Wait(1000)
    FormFeed
    LCD.tx(17)
    LCD.str(String("Starting..."))
    LCD.tx(211)                        'Startup melody
    LCD.tx(220)
    LCD.tx(224)
    LCD.tx(227)
    LED.init
    LED.LED(4000)                       'flash LEDs
    CustomChar
    Wait(1000)
    LED.LED(0)
    SamplingFreq := _rate                'initialize
    AlarmThresh := _thresh              'initialize
    Top

Pub Top
    WaitForNoButton
    repeat
        LCD.tx(18)

```

```

FormFeed
LCD.str(@TopMenu[i * 32])
LCD.tx(148)
LCD.str(String("  "))
LCD.tx(0)
LCD.tx(1)
LCD.str(String(" OK:Select"))

WaitForAnyButton
if (ina & $F == UP)
    i += 1
    if (i > TopMenuLen)
        i := 0
if (ina & $F == DWN)
    i -= 1
    if (i < 0)
        i := TopMenuLen
if (ina & $F == SEL)
    case i
        0: Monitor
        1: Recording
        2: Playback
        3: Settings
Wait(500)

Pub Monitor
adc.ADCEngineStart(1024, @maxValue)
WaitForNoButton
LCD.tx(18)
FormFeed
LCD.str(String("Monitoring:Alarm"))
LCD.tx(3)
LCD.str(String(": To Stop"))
Stopped := false
cognew(spinMonitor(@AlarmOn, @maxfreq, @maxamp, @Stopped), @Stack)
Wait(500)
LCD.tx(18)
repeat until (ina & BCK == BCK)
    if AlarmOn == true
        AlarmSignal
        AlarmOn := false
        FormFeed
        LCD.str(String("Monitoring:Alarm"))
        LCD.tx(3)
        LCD.str(String(": To Stop"))
        LED.LED(maxValue)
        maxValue := 0
        Wait(100)

    Stopped := true
    LED.LED(0)
    adc.ADCEngineStop
Top

Pub Recording
WaitForNoButton
repeat
    FormFeed
    LCD.tx(2)
    LCD.str(@RecMenu[j * 32])
    brb
    LCD.str(String(" OK:Select"))
    WaitForAnyButton
    if (ina & $F == UP)
        j += 1
        if (j > RecMenuLen)
            j := 0
    if (ina & $F == DWN)
        j -= 1
        if (j < 0)
            j := RecMenuLen

```

```

    if (ina & $F == BCK)
        Top
    if (ina & $F == SEL)
        case j
            0: TimedRec
            1: RecUntilStop
        Wait(500)

Pub TimedRec | temp
adc.ADCEngineStart(SamplingFreq, @maxvalue)
adc.FATEngineStart
WaitForNoButton
k:=2
repeat
    LCD.tx(18)
    FormFeed
    LCD.tx(2)
    LCD.tx(2)
    LCD.str(String("Record "))
    LCD.str(@RecLength[k * 7])
    brb
    LCD.str(String(" OK:Record"))
    LCD.tx(137)
    LCD.tx(24)
    WaitForAnyButton
    if (ina & $F == UP)
        k += 1
        if (k > RecValLen)
            k := 0
    if (ina & $F == DWN)
        k -- 1
        if (k < 0)
            k := RecValLen
    if (ina & $F == BCK)
        adc.unmount
        adc.ADCEngineStop
        adc.FATEngineStop
        Recording
    if (ina & $F == SEL)
        CardCheck := \adc.SDCardCheck
        if CardCheck <> true
            FormFeed
            LCD.str(String("No SD Card found"))
            Wait(2000)
        else
            Stopped := false
            RecTime := RecVal[k] * 60
            cognew(spinRecorder(@RecTime, @Stopped, @filename), @Stack)
            LCD.tx(18)
            FormFeed
            LCD.str(String("Recording "))
            LCD.str(@RecLength[k * 7])
            LCD.tx(148)
            LCD.str(String("Press OK to Stop"))
            Wait(4000)
            WaitForNoButton
            repeat until ((Stopped == true) OR (ina & SEL == SEL))
                LED.LED(maxValue)
                maxvalue := 0
                LCD.tx(128)
                LCD.str(String("Rec < "))
                if RecTime > 3600
                    LCD.dec((RecTime / 3600) + 1 )
                    LCD.str(String(" hr rem "))
                else
                    temp := RecTime / 60 + 1
                    LCD.dec(temp)
                    LCD.str(String(" m rem "))
                Wait(200)
            Stopped := true
            FormFeed

```

```

        LCD.str(String("Recording          Stopped"))
        LED.LED(0)
        Wait(3000)
        adc.unmount
        WaitForNoButton
        Wait(500)

Pub RecUntilStop
    adc.ADCEngineStart(SamplingFreq, @maxvalue)
    adc.FATEngineStart
    WaitForNoButton
    repeat
        LCD.tx(18)
        FormFeed
        LCD.tx(2)
        LCD.tx(2)
        LCD.str(String("Rec Continuous"))
        BackSym
        LCD.str(String(" OK: Start Rec"))
        waitpne(%0000,%1001,0)          'wait for BCK or SEL button to be pressed
        if (ina & $F == BCK)
            adc.ADCEngineStop
            adc.FATEngineStop
            Recording
        if (ina & $F == SEL)
            CardCheck := \adc.SDCardCheck
            if CardCheck <> true
                FormFeed
                LCD.str(String("No SD Card found"))
                Wait(2000)
            else
                Stopped := false
                RecTime := 0
                cognew(spinRecorder(@RecTime, @Stopped, @filename), @Stack)
                LCD.tx(18)
                FormFeed
                LCD.str(String("Recording... "))
                LCD.tx(148)
                LCD.str(String("Press OK to Stop"))
                WaitForNoButton
                repeat until ((Stopped == true) OR (ina & SEL == SEL))
                    LED.LED(maxValue)
                    maxValue := 0
                    Wait(200)
                Stopped := true
                LED.LED(0)
                FormFeed
                LCD.str(String("Recording          Stopped"))
                Wait(3000)
                WaitForNoButton
            Wait(500)

Pub Playback
    dac.FATEngineStart
    dac.DACEngineStart(5000)
    CardCheck := \adc.SDCardCheck
    if CardCheck <> true
        FormFeed
        LCD.str(String("No SD Card found"))
        Wait(2000)
        dac.unmount
        dac.FATEngineStop
        dac.DACEngineStop
        Top
    WaitForNoButton
    filename:=\dac.ListNextFile
    repeat
        FormFeed
        LCD.tx(2)
        LCD.str(filename)
        BackSym

```

```

LCD.tx(1)
LCD.str(String(" OK:Play"))
WaitForAnyButton
if (ina & $F == DWN)
    filename:=\dac.ListNextFile
if (ina & $F == BCK)
    dac.unmount
    dac.FATEngineStop
    dac.DACEngineStop
    Top
if (ina & $F == SEL)
    if filename == String("No WAV files")
        FormFeed
        LCD.str(String("Not a valid file"))
    else
        x10 := false
        FormFeed
        LCD.str(filename)
        LCD.tx(148)
        cognew(spinPlayer(filename,@Stopped, @x10), @Stack)
        Wait(500)
        WaitForNoButton
        repeat until ((Stopped == true) OR (ina & SEL == SEL))
            LCD.tx(148)
            LCD.tx(0)
            LCD.str(String(":"))
            if x10 == false
                LCD.dec(10)
            else
                LCD.dec(1)
            LCD.str(String("x OK:Stop "))
            if (ina & $F == UP)
                not x10
                WaitForNoButton
                Wait(200)

            Stopped := true
            FormFeed
            LCD.str(String("Playback Stopped"))
            Wait(2000)
            WaitForNoButton
            Wait(500)

```

Pub Settings

```

WaitForNoButton
repeat
    FormFeed
    LCD.tx(2)
    LCD.str(@SetMenu[j * 32])
    brb
    LCD.str(String(" OK:Select"))
    WaitForAnyButton
    if (ina & $F == UP)
        j += 1
        if (j > SetMenuLen)
            j := 0
    if (ina & $F == DWN)
        j -= 1
        if (j < 0)
            j := SetMenuLen
    if (ina & $F == BCK)
        Top
    if (ina & $F == SEL)
        case j
            0: SetSampFreq
            1: SetAlarmThresh
    Wait(500)

```

Pub SetSampFreq

```

WaitForNoButton

```

```

k:=1000
if (SamplingFreq <> 0)
  k := SamplingFreq
repeat
  FormFeed
  LCD.tx(2)
  LCD.tx(2)
  LCD.str(String("Freq: "))
  LCD.Dec(k)
  LCD.str(String(" Hz"))
  BackSym
  LCD.tx(0)
  LCD.tx(1)
  LCD.str(String(" OK:Set"))
  LCD.tx(24)
  LCD.tx(136)
  WaitForAnyButton
  if (ina & $F == UP)
    k += 1000
    if (k > 22000)
      k := 22000
  if (ina & $F == DWN)
    k -= 1000
    if (k < 1000)
      k := 1000
  if (ina & $F == BCK)
    Settings
  if (ina & $F == SEL)
    SamplingFreq := k
    FormFeed
    LCD.str(String("Frequency Set"))
    Wait(3000)
    WaitForNoButton
    Wait(500)

Pub SetAlarmThresh
WaitForNoButton
k:= AlarmThresh
repeat
  FormFeed
  LCD.tx(2)
  LCD.tx(2)
  LCD.str(String("Threshold "))
  LCD.Dec(k)
  LCD.str(String("%"))
  brb
  LCD.str(String(" OK:Set"))
  LCD.tx(24)
  LCD.tx(140)
  WaitForAnyButton
  if (ina & $F == UP)
    k += 10
    if (k > 90)
      k := 90
  if (ina & $F == DWN)
    k -= 10
    if (k < 10)
      k := 10
  if (ina & $F == BCK)
    Settings
  if (ina & $F == SEL)
    AlarmThresh := k
    FormFeed
    LCD.str(String("Threshold Set"))
    Wait(3000)
    WaitForNoButton
    Wait(500)

Pub CustomChar

LCD.tx(248)                                     ' Define custom character 0 (Up arrow)

```

```

LCD.tx(%00100)
LCD.tx(%01110)
LCD.tx(%11111)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)

LCD.tx(249)                                ' Define custom character 1 (Down arrow)

LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%00100)
LCD.tx(%11111)
LCD.tx(%01110)
LCD.tx(%00100)

LCD.tx(250)                                ' Define custom character 2 (Right arrow)

LCD.tx(%00000)
LCD.tx(%00100)
LCD.tx(%00110)
LCD.tx(%11111)
LCD.tx(%00110)
LCD.tx(%00100)
LCD.tx(%00000)
LCD.tx(%00000)

LCD.tx(251)                                ' Define custom character 3 (Left arrow)

LCD.tx(%00000)
LCD.tx(%00100)
LCD.tx(%01100)
LCD.tx(%11111)
LCD.tx(%01100)
LCD.tx(%00100)
LCD.tx(%00000)
LCD.tx(%00000)

Pub AlarmSignal
'Displays the detected amplitude and frequency and plays tones, flashes screen
LCD.tx(17)
FormFeed
LCD.Str(String("! "))
LCD.dec(maxamp)
LCD.Str(String(" @ "))
LCD.dec(maxfreq)
LCD.Str(String(" Hz !"))
LCD.tx(212)
AlarmOn := false
repeat 15
  LCD.tx(225)
  LCD.tx(220)
  Wait(500)
  LCD.tx(18)
  LED.LED(0)
  Wait(500)
  LCD.tx(17)
  LED.LED(4000)
  if (ina & $F == BCK)
    quit
WaitForNoButton
LCD.tx(18)

PUB spinRecorder(RecTimeAddr, StoppedAddr, filenameaddr) ' Starts the recording

  adc.startRecordingWAVFile(RecTimeAddr, StoppedAddr, filenameaddr)

```

```

    cogstop(cogID)

PUB spinMonitor(AlarmOnAddr, maxfreqaddr, maxampaddr, StoppedAddr) ' Starts monitoring

    adc.Monitor(AlarmThresh, AlarmOnAddr, maxfreqaddr, maxampaddr, StoppedAddr)
    cogstop(cogID)

PUB spinPlayer(name, StoppedAddr, x10addr) ' Starts playback

    dac.startPlayingWAVFile(name, StoppedAddr, x10addr)
    cogstop(cogID)

PUB brb 'bottom row buttons

    BackSym
    LCD.tx(0)
    LCD.tx(1)

PUB BackSym '2nd line of display, space, then back symbol

    LCD.tx(148)
    LCD.tx(9)
    LCD.tx(3)

PUB FormFeed

    LCD.tx(12)
    Wait(10)
    LCD.tx(22)

PUB Wait(length) ' pause execution for 'length' msecs

    waitcnt((clkfreq / 1000 * length) + cnt)

PUB WaitForNoButton

    waitpeq(%0000,%1111,0) 'wait until no buttons are pressed

PUB WaitForAnyButton

    waitpne(%0000,%1111,0) 'wait for any button to be pressed

DAT
TopMenu      byte      "Monitor w/ Alarm",0[16]
             byte      "Recording",0[23]
             byte      "Playback",0[24]
             byte      "Settings",0[24]
TopMenuLen   byte      3
RecMenu      byte      "Timed Recording", 0[17]
             byte      "Rec until Stop",0[18]
RecMenuLen   byte      1
SetMenu      byte      "Sampling Freq", 0[19]
             byte      "Alarm Threshold", 0[17]
SetMenuLen   byte      1
RecLength    byte      "1 min",0[2]
             byte      "10 min",0[1]
             byte      "30 min",0[1]
             byte      "1 hr",0[3]
             byte      "2 hr",0[3]
             byte      "4 hr",0[3]
             byte      "8 hr",0[3]
             byte      "12 hr",0[2]
             byte      "24 hr",0[2]
RecValLen    byte      8
RecVal       word      1,10,30,60,120,240,480,720,1440
BCK          byte      %0001
UP           byte      %0010
DWN         byte      %0100
SEL         byte      %1000

{{

```



```

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/////////////////////////////////////////////////////////////////
}}

```

Heater_fft.spin

```

' heater_fft module
' This module performs a 1024 point fft and returns the results in the first 512 cells of
' the datablock.
' Original Module "heater_fft" found on Parallax Object Exchange (OBEX) obex.parallax.com

' In place Radix-2 Decimation In Time FFT
'
' Michael Rychlik. 2011-1-25
'
'User optimization controls
'#define PASM_BUTTERFLIES      'Set this for fast PASM FFT, about 30ms
'#define USE_FASTER_MULT      'Set this for faster multiply
'#define USE_FASTER_SQRT      'Set this for faster but much bigger square root.

```

```

-----
CON
'Specify size of FFT buffer here with length and log base 2 of the length.
'N.B. Changing this will require changing the "twiddle factor" tables.
' and may also require changing the fixed point format (if going bigger)
FFT_SIZE      = 1024
LOG2_FFT_SIZE = 10

CMD_DECIMATE  = %0001
CMD_BUTTERFLY = %0010
CMD_MAGNITUDE = %0100
CMD_TEST     = %1000

```

```

-----
VAR
    long mailboxp
    byte cog

```

```

-----
PUB start (mailp)
'#ifdef PASM_BUTTERFLIES
    mailboxp := mailp
    LONG[mailboxp] := 0
    cog := cognew (@bfly, mailp)    'Check error?

```

```

Pub stop

    cogstop(cog)

```

```

PUB butterflies(cmd, bxp, byp)
    LONG[mailbox + 4] := bxp          'Address of x buffer
    LONG[mailbox + 8] := byp         'Address of y buffer
    LONG[mailbox + 0] := cmd         'Do butterflies and/or decimation
    repeat while LONG[mailbox + 0] <> 0
-----
-----
DAT
bfly      org      0
          mov      mb_ptr, par
          rdlong   command, mb_ptr wz      'Wait for run command in mailbox
          if_z     jmp      #bfly

          add      mb_ptr, #4
          rdlong   bx_ptr, mb_ptr          'Fetch x array address from mbox

          add      mb_ptr, #4
          rdlong   by_ptr, mb_ptr          'Fetch y array address from mbox
          sub      mb_ptr, #8

          test     command, #CMD_DECIMATE wz  'Bit reversal required on data?
          if_z     jmp      #:no_decimate

'Radix-2 decimation in time. (The bit reversal stage)
'Moves every sample of bx to a position given by reversing the bits of its original array
'index. This is a direct translation of the Spin decimate above, original Spin code used
'as comments. N.B. Only the x array is bit-reversed it is up to the app to clear y.

          mov      c, fft_size_
          mov      b, #0
          'repeat i from 0 to FFT_SIZE - 1

:dloop    mov      a, b
          mov      rev_a, a
          rev      rev_a, #32 - LOG2_FFT_SIZE
          'revi := i >< LOG2_FFT_SIZE

          cmp      a, rev_a wz
          if_nc    jmp      #:skip_rev
          'if i < rev_i

          shl      a, #2
          shl      rev_a, #2
          'Times 4 as we are reading longs

          mov      hub_ptr, bx_ptr
          add      hub_ptr, a
          rdlong   tx, hub_ptr
          'tx1 := long[bxp + i * 4]

          mov      hub_rev_ptr, bx_ptr
          add      hub_rev_ptr, rev_a
          rdlong   ty, hub_rev_ptr
          wrlong   ty, hub_ptr
          'long[bxp + i * 4] := long[bxp + rev_i * 4]

          wrlong   tx, hub_rev_ptr
          'long[bxp + rev_i * 4] := tx1

:skip_rev add      b, #1
          djnz     c, #:dloop

:no_decimate
          test     command, #CMD_BUTTERFLY wz  'Perform buterflies?
          if_z     jmp      #:no_butterfly

'Apply FFT butterflies to N complex samples in buffers bx and by, in time decimated order
'Resulting FFT is produced in bx and by in the correct order.
'This is a direct translation from the Spin code above, original Spin code in comments.

          mov      flight_max, fft_size_
          sar      flight_max, #1
          mov      wangleSkip, fft_size_
          shl      wangleSkip, #2
          'flight_max := FFT_SIZE / 2
          'wangleSkip := FFT_SIZE * 4

          mov      butterflySpan, #4
          'butterflySpan := 4

```

```

mov      butterfly_max, #1          'butterfly_max := 1
mov      flightSkip, #4            'flightSkip := 4

'Loop through all the decimation levels
:lloop  mov      level, #LOG2_FFT_SIZE  'level := LOG2_FFT_SIZE
        'repeat
        mov      b0x_ptr, bx_ptr      'b0x_ptr := @bx
        mov      b0y_ptr, by_ptr      'b0y_ptr := @by

        mov      blx_ptr, b0x_ptr      'blx_ptr := b0x_ptr + butterflySpan
        add      blx_ptr, butterflySpan

        mov      bly_ptr, b0y_ptr      'bly_ptr := b0y_ptr + butterflySpan
        add      bly_ptr, butterflySpan

'Loop though all the flights in a level
:floop  mov      flight, flight_max    'flight := flight_max
        'repeat
{new}   mov      wangle, #0

'Loop through all the butterflies in a flight
        mov      butterfly, butterfly_max  'butterfly := butterfly_max

'Do the initial pass optimization, when W = [1,0] we don't need to multiply
' c = 1 (well, 4096/4096), d = 0
        mov      k2, #0                'k2 := (d * (a + b)) / 4096
        rdlong   a, blx_ptr             'a := LONG[blx_ptr]
        mov      k1, a                  'k1 := (a * (c + d)) / 4096
        neg      k3, a                  'k3 := (c * (b - a)) / 4096
        rdlong   b, bly_ptr             'b := LONG[bly_ptr]
        add      k3, b                  'k3 := (c * (b - a)) / 4096
        jmp      #:continue_bloop

:bloop  ' repeat                          'At last...the butterfly.
        rdlong   a, blx_ptr             'a := LONG[blx_ptr]

        'Precompute the optimization for c=0, d=-1
        neg      k1, a                  'k1 := (a * (c + d)) / 4096
        neg      k2, a                  'k2 := (d * (a + b)) / 4096

        rdlong   b, bly_ptr             'b := LONG[bly_ptr]

        'Precompute the optimization for c=0, d=-1
        sub      k2, b                  'k2 := (d * (a + b)) / 4096
        mov      k3, #0                 'k3 := (c * (b - a)) / 4096

{getcos}  mov      c, wangle
        add      c, sin_90              'For cosine, add 90°
        test     c, sin_90              wc   'Get quadrant 2|4 into c
        test     c, sin_180            wz   'Get quadrant 3|4 into nz
        negc     c, c                   'If quadrant 2|4, negate offset
        or      c, sin_table           'OR in sin table address >> 1
        shl     c, #1                  'Shift left to get final word address
        rdword   c, c                   'Read word sample from $E000 to $F000
        negnz   c, c                   'If quadrant 3|4, negate sample

        sar     c, #4 wz                'Scale to +/- 4095

        if_z    jmp     #:continue_bloop ' if c==0, we already have k1, k2, k3

{getsin}  mov      d, wangle
        test     d, sin_90              wc   'Get quadrant 2|4 into c
        test     d, sin_180            wz   'Get quadrant 3|4 into nz
        negc     d, d                   'If quadrant 2|4, negate offset
        or      d, sin_table           'OR in sin table address >> 1
        shl     d, #1                  'Shift left to get final word address
        rdword   d, d                   'Read word sample from $E000 to $F000
        negnz   d, d                   'If quadrant 3|4, negate sample

        sar     d, #4                  'Scale to +/- 4095
        neg     d, d                    'We want -cos

```

```

mov     m1, c                'k1 := (a * (c + d)) / 4096
add     m1, d
mov     m2, a
call    #mult
mov     k1, m1
sar     k1, #15 - 3

mov     m1, a                'k2 := (d * (a + b)) / 4096
add     m1, b
mov     m2, d
call    #mult
mov     k2, m1
sar     k2, #15 - 3

mov     m1, b                'k3 := (c * (b - a)) / 4096
sub     m1, a
mov     m2, c
call    #mult
mov     k3, m1
sar     k3, #15 - 3

:continue_bloop

mov     tx, k1                'tx := k1 - k2 (part I)
mov     ty, k1                'ty := k1 + k3 (part I)

rdlong  k1, b0x_ptr           'k1 := LONG[b0x_ptr]

sub     tx, k2                ' (part II) moved from above to take
add     ty, k3                ' advantage of the hub wait times

rdlong  k2, b0y_ptr           'k2 := LONG[b0y_ptr]

mov     a, k1                 'LONG[blx_ptr] := k1 - tx
sub     a, tx
wrlong  a, blx_ptr

mov     a, k2                 'LONG[bly_ptr] := k2 - ty
sub     a, ty
wrlong  a, bly_ptr

mov     a, k1                 'LONG[b0x_ptr] := k1 + tx
add     a, tx
wrlong  a, b0x_ptr

mov     a, k2                 'LONG[b0y_ptr] := k2 + ty
add     a, ty
wrlong  a, b0y_ptr

add     b0x_ptr, #4           'b0x_ptr += 4
add     b0y_ptr, #4           'b0y_ptr += 4

add     blx_ptr, #4           'blx_ptr += 4
add     bly_ptr, #4           'bly_ptr += 4

add     wangle, wangleSkip    'wangle += wangleSkip

djnz    butterfly, #:bloop    'while --butterfly <> 0

add     b0x_ptr, flightSkip    'b0x_ptr += flightSkip
add     b0y_ptr, flightSkip    'b0y_ptr += flightSkip
add     blx_ptr, flightSkip    'blx_ptr += flightSkip
add     bly_ptr, flightSkip    'bly_ptr += flightSkip
djnz    flight, #:floop       'while --flight <> 0

shl     butterflySpan, #1      'butterflySpan <=<= 1
shl     flightSkip, #1         'flightSkip <=<= 1

shr     flight_max, #1         'flight_max >>= 1

```

```

        shr        wangleSkip, #1
        shr        wSkip, #1                'wSkip >= 1
        shl        butterfly_max, #1        'butterfly_max <= 1
        djnz       level, #:lloop           'while --level <> 0
:no_butterfly
        test       command, #CMD_MAGNITUDE wz 'Calculate magnitudes?
        if_z      jmp        #:no_magnitude

'Calculate magnitudes from the complex results in x and y. Results placed into x

        mov        c, fft_size_             'repeat i from 0 to FFT_SIZE
        add        c, #1                    'That is one more than half FFT_SIZE
                                                'so as to include the Nyquist freq

        mov        b0x_ptr, bx_ptr
        mov        b0y_ptr, by_ptr

:mloop
        rdlong     m1, b0x_ptr
        sar        m1, #LOG2_FFT_SIZE - 1
        mov        m2, m1
        call       #mult
        mov        input, m1

        rdlong     m1, b0y_ptr
        sar        m1, #LOG2_FFT_SIZE - 1
        mov        m2, m1
        call       #mult
        add        input, m1

        call       #sqrt

        wrlong     root, b0x_ptr            'Write result to x array

        add        b0x_ptr, #4              'Next x and y element and loop
        add        b0y_ptr, #4
        djnz       c, #:mloop

:no_magnitude
        mov        command, #0
        wrlong     command, mb_ptr
        jmp        #bfly

-----

mult      'Account for sign
#ifdef USE_FASTER_MULT
        abs        m1, m1 wc
        negc       m2, m2
        abs        m2, m2 wc
        'Make t2 the smaller of the 2 unsigned parameters
        mov        m3, m1
        max        m3, m2
        min        m2, m1
        'Correct the sign of the adder
        negc       m2, m2
{{#else
        abs        m3, m1 wc
        negc       m2, m2
#endif}}

        'My accumulator
        mov        m1, #0
        'Do the work
:mul_loop shr        m3, #1 wc,wz           'Get the low bit of t2
        if_c      add        m1, m2         'If it was a 1, add adder to accumulator
        shl        m2, #1                  'Shift the adder left by 1 bit
        if_nz     jmp        #:mul_loop     'Continue as long as there are no more 1's
mult_ret  ret

m1        long      0
m2        long      0
m3        long      0
-----

```

```

-----
#ifdef USE_FASTER_SQRT
Faster code square root (Chip Gracey after discussion with lonesock on Propeller Forums)
sqrt      mov      root, h40000000
          cmpsub   input, root wc
          sumnc    root, h40000000
          shr      root, #1

          or       root, h10000000
          cmpsub   input, root wc
          sumnc    root, h10000000
          shr      root, #1

          or       root, h04000000
          cmpsub   input, root wc
          sumnc    root, h04000000
          shr      root, #1

          or       root, h01000000
          cmpsub   input, root wc
          sumnc    root, h01000000
          shr      root, #1

          or       root, h00400000
          cmpsub   input, root wc
          sumnc    root, h00400000
          shr      root, #1

          or       root, h00100000
          cmpsub   input, root wc
          sumnc    root, h00100000
          shr      root, #1

          or       root, h00040000
          cmpsub   input, root wc
          sumnc    root, h00040000
          shr      root, #1

          or       root, h00010000
          cmpsub   input, root wc
          sumnc    root, h00010000
          shr      root, #1

          or       root, h00004000
          cmpsub   input, root wc
          sumnc    root, h00004000
          shr      root, #1

          or       root, h00001000
          cmpsub   input, root wc
          sumnc    root, h00001000
          shr      root, #1

          or       root, h00000400
          cmpsub   input, root wc
          sumnc    root, h00000400
          shr      root, #1

          or       root, #$100
          cmpsub   input, root wc
          sumnc    root, #$100
          shr      root, #1

          or       root, #$40
          cmpsub   input, root wc
          sumnc    root, #$40
          shr      root, #1

          or       root, #$10
          cmpsub   input, root wc

```

```

        sumnc    root, #10
        shr     root, #1

        or      root, #4
        cmpsub  input,root wc
        sumnc   root, #4
        shr     root, #1

        or      root, #1
        cmpsub  input,root wc
        sumnc   root, #1
        shr     root, #1
sqrt_ret    ret

h10000000  long    $10000000
h04000000  long    $04000000
h01000000  long    $01000000
h00400000  long    $00400000
h00100000  long    $00100000
h00040000  long    $00040000
h00010000  long    $00010000
h00004000  long    $00004000
h00001000  long    $00001000
h00000400  long    $00000400

{{#else

'Faster code square root (Chip Gracey after discussion with lonesock on Propeller Forums)
sqrt      mov     root, #0                'Reset root
          mov     mask, h40000000        'Reset mask (constant in register)
:sqloop   or      root, mask              'Set trial bit
          cmpsub  input, root wc         'Subtract root from input if fits
          sumnc   root, mask             'Cancel trial bit, set root bit if fit
          shr     root, #1                'Shift root down
          shr     mask, #2                'Shift mask down
          tjnz   mask, #:sqloop          'Loop until mask empty
sqrt_ret  ret
#endif }}
h40000000  long    $40000000
'-----

'-----
'Large constants
fft_size_  long    FFT_SIZE
sin_90     long    $0800
sin_180    long    $1000
sin_table  long    $E000 >> 1          'ROM sin table base shifted right

'COG variables
level      long    0
flight     long    0
butterfly  long    0
flight_max long    0
wSkip      long    0
butterflySpan long  0
butterfly_max long  0
flightSkip long    0
k1         long    0
k2         long    0
k3         long    0
a          long    0
b          long    0
c          long    0
d          long    0
tx         long    0
ty         long    0
b0x_ptr    long    0
b0y_ptr    long    0
blx_ptr    long    0
bly_ptr    long    0
mb_ptr     long    0

```

```

bx_ptr      long      0
by_ptr      long      0
wangle      long      0
wangleSkip  long      0

rev_a       long      0
hub_ptr     long      0
hub_rev_ptr long      0
command     long      0
root        long      0
mask        long      0
input       long      0

```

```

-----
                fit      496
#endif
-----

```

```

-----
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,
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-----

```

LED Driver.spin

```

{{ LED Driver
\\ This module initializes communication with the LED driver, as well as takes an input
\\ value, decides the LEDs to turn on and send the command to the LED driver.
}}
CON
    ValMax = 4095
    CLK = 5
    SDI = 4
    LE = 6
    OE = 7

OBJ

PUB init

    dira[CLK]~~
    dira[SDI]~~
    dira[LE]~~
    dira[OE]~~
    outa[OE] := 0
    outa[CLK] := 0
    outa[LE] := 0

PUB LED(Value) | data

    data := %00000000
    if Value > 2457

```



```

// WAV-Player - Author: Kwabena W. Agyeman
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
}}
CON
    dopin = 12
    clkpin = 13
    dipin = 11
    cspin = 8
    cdpin = -1 ' -1 if unused.
    wppin = -1 ' -1 if unused.

    rtcres1 = -1 ' -1 always.
    rtcres2 = -1 ' -1 always.
    rtcres3 = -1 ' -1 always.
OBJ fat: "SD-MMC_FATEngine.spin"

VAR

    long clocksPerSample, samplesPerSecond, dataSize, fileSize
    word callerPointer, callePointer, data2DAC, count, LastVal
    byte stopped, cogIdentification
    long writeResult, curpos, i, switch, linecount, playFlag
    word datablock[512]

PUB startPlayingWAVFile(filePathName, StoppedAddr, x10addr) '' 52 Stack Longs
    result := \PlayWAV(filePathName, StoppedAddr, x10addr)

PRI PlayWAV(filePathName, StoppedAddr, x10addr) | x10' 48 Stack Longs
    ifnot(fat.partitionMounted)
        fat.mountPartition(0)
    result := fat.openFile((filePathName), "R")
    fat.fileSeek(4)
    fileSize := fat.readLong + 8
    fat.fileSeek(24) 'get samples per second
    samplesPerSecond := fat.readLong
    fat.fileSeek(40) 'get number of data bytes
    dataSize := fat.readLong
    clocksPerSample := (clkfreq / samplesPerSecond) 'calculate clocks to wait per samp
    fat.fileSeek(44) 'move to first data chunk
    stopped := false
    curpos := 0

    fat.fileSeek(44)
    callerPointer := callePointer := 0 'init flag
    fat.readData(@datablock,512)
    not callerPointer ' callerpointer = -1
    'callePointer was set to 0. After a block of DAC PASM, it will flip to -1.
    'Wait until it equals 0 again (playing A block)
    'then continue loading data into block B (callerpointer = -1)
    repeat while (callerPointer <> callepointer) 'callerpointer=1, callepointer = 0
    repeat while(callerPointer == callePointer) 'callerpointer=1, callepointer = 1
    long[StoppedAddr] := false 'ready to play first block
    repeat until(long[StoppedAddr] == true)
        x10 := long[x10addr]
        if x10 == true
            clocksPerSample := ((clkfreq / samplesPerSecond) / 10) #> (clkfreq / 80000)
        else
            clocksPerSample := (clkfreq / samplesPerSecond)
        playflag := 1
        fat.readData(@dataBlock[256 & callerPointer], 512) 'read next data block
        not callerPointer
        repeat while (callerpointer <> callepointer)
        curpos := fat.filetell
        if curpos >= fileSize - 512
            fat.fileseek(44) 'loops playback
    playflag := 0

PUB DACEngineStart(sampleRate) '' 9 Stack Longs

'' // Starts up the ADC driver running on a cog.

```



```

if_nz                wrword  buffer,      callePointerAddress  '
                    mov     playerPointer, dataBlockAddress  '

                    jmp     #outerLoop  wz          ' Loop.

' ///////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
'                               Data
' ///////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////

mask12               long    $FFF
dmask                long    1 << 15
cmask                long    1 << 14
smask                long    1 << 10
comInit              word    $3000
shutdown             long    $1000
command              long    0
value                word    0

' ////////////////////////////////////////////////////////////////////Addresses/////////////////////////////////////////////////////////////////

clocksPerSampleAddress long  0
dataBlockAddress       long  0
callePointerAddress    long  0
playFlagAddress        long  0

' ////////////////////////////////////////////////////////////////////Run Time Variables/////////////////////////////////////////////////////////////////

buffer                res    1
counter               res    1
playerPointer         res    1
playerRate            res    1
timeCounter           res    1
bits                  res    1
outputOn              res    1

' ///////////////////////////////////////////////////////////////////

fit                   496

CON WAVFileHeaderSize = 44 ' DO NOT EDIT!

DAT WAVFileHeaderData ' DO NOT EDIT!

' ///////////////////////////////////////////////////////////////////

byte byte "RIFF" ' "RIFF" chunk header.
byte long 0 ' "RIFF" chunk size = (fileSize - 8). Offset 4.
byte byte "WAVE" ' File type.

byte byte "fmt " ' "fmt " chunk header.
byte long 16 ' "fmt " chunk size.
byte word 1 ' Audio format.
byte word 1 ' Nuber of channels.
byte long 0 ' Sample rate.EDITED by setup function!
byte long 0 ' Byte rate. EDITED by setup function!
byte word 2 ' Block align.
byte word 16 ' Bits per sample.

byte byte "data" ' "data" chunk header.
byte long 0 ' "data" chunk size = (fileSize - 44). Offset 40.

' ///////////////////////////////////////////////////////////////////

{{

/////////////////////////////////////////////////////////////////
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```



```

' //////////////////////////////////////////////////////////////////Inner Loop////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
innerLoop          waitcnt timeCounter,      recorderRate      'wait for next sample
main_loop          mov      command,#$10          'init command
                  mov      t2,enables          'get enables
                  mov      t3,#8              'ready 8 channels
cloop              shr      t2,#1              wc              'if channel disabled, skip
                  test     t2,#$80          wc              'channel enabled, get single/diff
                  muxnc   command,#$08
                  mov      stream,command
                  or       outa,smask          'CS high
                  or       dira,dmask        'make DIN/DOUT output
                  mov      bits,#20          '(cs+1+diff+ch[3]+0+0+data[12])
bloop              test     stream,$$20      wc              'update DIN/DOUT
                  muxc   outa,dmask
if_z               cmp      bits,#14          wz              'if command done, input DIN/DOUT
                  andn   dira,dmask
                  andn   outa,cmask          'CLK low
                  or     outa,cmask          'CLK high
                  test     dmask,ina         wc              'sample DIN/DOUT
                  rcl    stream,#1
                  andn   outa,smask          'CS low
                  djnz   bits,#bloop        'next data bit
                  and    stream,mask12      'trim and write sample
                  wrword  stream,recorderPointer 'write data word to memory
                  rdlong  maxval, maxValueAddress 'get current maximum value
                  cmp     maxval, stream wc
if_c               wrlong  stream, maxValueAddress 'if stream > current max val
                  shr     stream, #1        'bx values < 2048 so divide by 2
                  wrlong  stream,bxptr      'write value to bx buffer for fft
                  add     recorderPointer, #2 'increment buffer pointers
                  add     bxptr, #4
                  rdlong  recorderRate,     clocksPerSampleAddress ' Loop.
                  djnz   counter,          #innerLoop
' //////////////////////////////////////////////////////////////////Outer Loop////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
if_nz              rdword  buffer,          callePointerAddress wz ' Switch pntr
                  sumz   buffer,          #1
                  wrword  buffer,          callePointerAddress
if_z               mov     recorderPointer, dataBlockAddress
                  sub    bxcounter, #1    wz
if_z               call   #bxbuffer
                  jmp    #outerLoop      ' Loop.
bxbuffer           rdword  buffer,          bxcallePointerAddress wz 'flip bxpntr
                  sumz   buffer,          #1
                  wrword  buffer,          bxcallePointerAddress
if_nz              mov     bxptr,          bxAddress
bxbuffer_ret       mov     bxcounter,      #4
                  ret
                  'reset bx cnt
                  '1024, 4 sets
                  'of 256)
' //////////////////////////////////////////////////////////////////
' Data

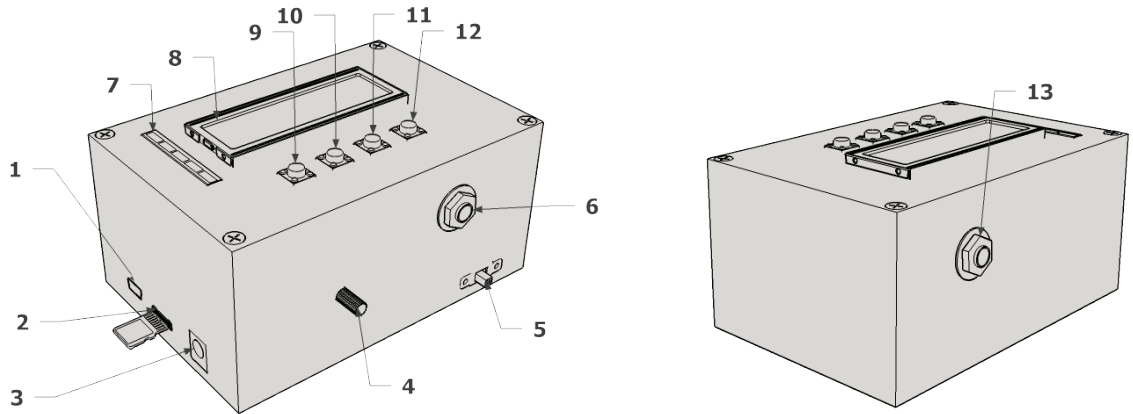
```



```
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////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
}}
```

FullDuplexSerial.spin - Unmodified code from Parallax, available in Propeller Tool 1.3.2

Infrasound Detection Device



1. Mini USB Connector – Used to program and/or power the device.
2. Micro SD Card Slot – Cards are inserted face down (metal contacts face up)
3. 2.1 mm Power Jack – Used to power device (6-9 VDC)
4. Gain knob – Controls amplifier gain (2-229x)
5. Power Switch
6. Output Jack – 3.5 mm mono output *** RANGE SPECS ONCE FINALIZED***
7. LED Strip – Displays input amplitude
8. Display – 2-line interface displays device instructions and status
9. Back button – Interface control – displayed as ←
10. Up Button – Interface control – displayed as ↑
11. Down Button – Interface control – displayed as ↓
12. Select Button – Interface control – displayed as **OK**
13. Input Jack – 3.5 mm mono input (0-2.5 VAC)

To operate device:

1. Slide power switch to ON position.
2. Display will illuminate showing `Starting...` and plays a startup tone. LED block will flash on and off.
3. Menu options will show:

The display has two lines. The top line will show information and status. The bottom line will show what button options are available. Press the UP and DOWN buttons to switch between the options. Press BACK in any menu to return to a higher-level menu. Press OK to select the option shown.

Menu Options

1. Monitor w/ Alarm
Press OK to select. This will put the device in Monitor mode. If the signal from the input is a frequency below 40 Hz that exceeds the Alarm Threshold level, the alarm will sound and display lights will flash. Pressing BACK will return the user to the main menu and disable the monitor/alarm.
2. Recording
Press OK to enter the Recording menu. Recordings are made in the WAV format. Filenames are auto-generated. The default filename is `000_OKC.WAV`. Subsequent files will increment the three-digit number at the front of the filename. Two options are available:
 - a. Timed Recording
Press OK to select. This option allows the user to record the input signal for a specified amount of time. Pressing UP or DOWN in this menu switches the predefined time options ranging from 1 minute to 24 hours. When the desired duration is displayed, press OK to start the recording. The device will record until the time has elapsed or maximum file size has been reached, whichever is earlier.
 - b. Rec Until Stop
Press OK to select. The display will show `Rec Continuous`. This option can start a recording that will continue until either the OK button is pressed or maximum file size is reached, whichever is earlier. Press OK to start and stop the recording.
3. Playback
Press OK to enter the Playback menu. Valid WAV file names will be shown on the display. Press DOWN to step through the files on the SD card. Press OK to start

playback of the selected file. While the file is playing, press DOWN to toggle between 1x and 10x playback speeds. Press OK to stop playback.

4. Settings

Press OK to enter the Settings menu. Two options are available:

a. Sampling Freq

Press OK to select. This option allows setting the sampling frequency between 1000 Hz and 20000 Hz in 1000 Hz increments. Press UP and DOWN to switch between frequency settings. Press OK to set the frequency. This value is only used while the device is powered on. On reset, the sampling frequency reverts to default (1000 Hz.)

b. Alarm Threshold

Press OK to select. This option allows setting the amplitude threshold that will trigger the alarm for frequencies below 40 Hz. The valid range is between 10% and 90% of maximum amplitude. Press UP and DOWN to switch between threshold percentages in increments of 10%. Press OK to set the threshold level. This value is only used while the device is powered on. On reset, the alarm threshold value reverts to default (10%.)

Setting Input Levels

When monitoring or recording, the gain of the amplifier must be adjusted to account for different input levels and different input devices.

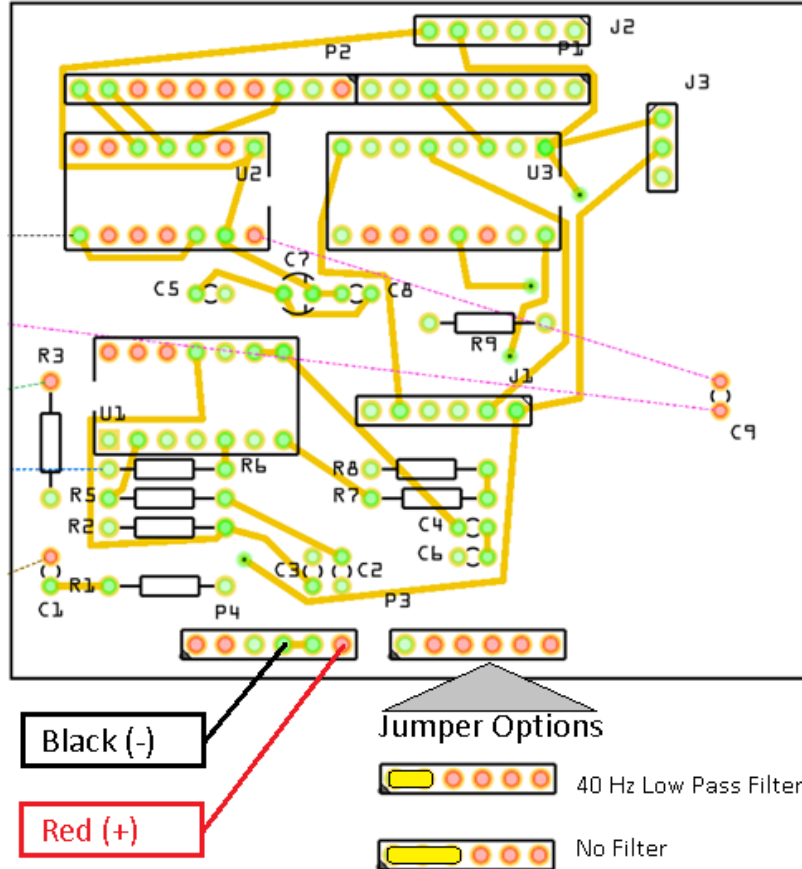
1. Put the device in Monitoring w/ Alarm or a Recording mode.
2. While in these modes, the 5-light LED bar will show the input signal level.
3. Provide an input signal of a type similar to the expected volume level of the recording. For example, if the device is a microphone and the expected volume level is similar to human speech, speak into the microphone at the approximate volume level.
4. Adjust the gain control knob while providing an input signal until the signal is consistently in the 3rd or 4th LED from the bottom. Reduce the gain if the 5th LED is lit constantly or often to prevent clipping the signal.
5. The gain level is now set.

Changing Batteries

Remove the 4 screws on the top cover. Carefully lift off the cover. The battery compartment is on the right face of the device. The device requires four AAA batteries.

Connections

There are three connections that could become unconnected if the cover is removed. Two of the connections are the + and – connections to the battery. The last connection is a jumper that connects the input signal to the microcontroller. The first option will connect through a low-pass filter (needed for infrasound detection.) The second option bypasses the low pass filter. Only one of these options can be connected at the same time. Ensure connections are in place before reattaching the top cover. Connections are made by pushing the jumper contact firmly into the female header slot (see figure below.)



VITA

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